## Steady state tracking

Recall that integral control gives zero steady state error to a step even in the presence of plant modelling mismatch.

This does not happen in our state feedback reference tracking scheme.

$$
u(k)=K\left(x_{r e f}-x(k)\right),
$$

so if $x(k) \longrightarrow x_{r e f}$ then $u(k) \longrightarrow 0$.
However with $u(k) \longrightarrow 0$, and no plant poles at $z=1$, we have, $y(k) \longrightarrow 0$. Clearly then, $y(k) \neq r(k)$, the reference.

## Feedforward compensation

Recall that the matrix $N_{u}$ can provide some steady-state feedforward compensation:

$$
u(k)=K\left(x_{r e f}-x(k)\right)+N_{u} r(k),
$$

where,

$$
N_{u}=\frac{1}{\text { plant steady state gain }} .
$$

This is only correct if we know the plant steady state gain.

## Integral control

Recall the idea behind integral control. If we consider the integral of the error at time, $t$,

$$
\int_{0}^{t}(r(\tau)-y(\tau)) d \tau
$$

we want to make this quantity go zero. In other words,

$$
\lim _{t \rightarrow \infty} \int_{0}^{t}(r(\tau)-y(\tau)) d \tau=0
$$

This means that as $t \longrightarrow \infty$, the error must go to zero,

$$
y(t) \longrightarrow r(t) .
$$

If this wasn't the case (i.e. in steady state $y \neq r$ ), then the integral would end up going to $\infty$.

## Approach

Make the integral of the error $(r(k)-y(k))$ go to zero for state feedback.

## Integral control



State space:

$$
\begin{aligned}
x(k+1) & =A x(k)+B u(k) \\
y(k) & =C x(k)
\end{aligned}
$$

Augment the output, $y(k)$ with an integrator.


## Augmented system

$$
\left[\begin{array}{c}
x_{I}(k+1) \\
x(k+1)
\end{array}\right]=\left[\begin{array}{cc}
I & C \\
0 & A
\end{array}\right]\left[\begin{array}{c}
x_{I}(k) \\
x(k)
\end{array}\right]+\left[\begin{array}{l}
0 \\
B
\end{array}\right] u(k) .
$$

This now has $n+1$ states (or, in general, $n$ plus the number of outputs).

## Applying state feedback

If we now design a state feedback controller (using pole placement),

$$
u(k)=-\left[\begin{array}{ll}
K_{I} & K
\end{array}\right]\left[\begin{array}{c}
x_{I}(k) \\
x(k)
\end{array}\right],
$$

then this will make $x(k) \longrightarrow 0$ and the integrator output go to zero.

$$
x_{I}(k)=\frac{1}{z-1} y(k) \longrightarrow 0 .
$$

## Reference tracking

Recall that we replaced $u(k)=-K x(k)$ by $u(k)=-K\left(x_{r e f}-x(k)\right)$.

## Key idea

Replace $x(k)$ by $x(k)-x_{r e f} \quad$ (state error)
and
$\frac{1}{z-1} y(k) \quad$ by $\quad \frac{1}{z-1}(y(k)-r(k)) \quad$ (integral of the tracking error).
This will make both the state error and the integral of the tracking error to go to zero.

## Implementation:

$$
u(k)=-\left[\begin{array}{ll}
K_{I} & K
\end{array}\right]\left[\begin{array}{c}
\frac{1}{z-1}(y(k)-r(k)) \\
x(k)-x_{r e f}
\end{array}\right]=\left[\begin{array}{ll}
K_{I} & K
\end{array}\right]\left[\begin{array}{c}
\frac{1}{z-1}(r(k)-y(k)) \\
x_{r e f}-x(k)
\end{array}\right] .
$$

## Implementation

$$
u(k)=\left[\begin{array}{ll}
K_{I} & K
\end{array}\right]\left[\begin{array}{c}
\frac{1}{z-1}(r(k)-y(k)) \\
x_{r e f}-x(k)
\end{array}\right] .
$$



If $x(k)$ is not measured: build an estimator and use $\hat{x}(k)$ in the above.
A feedforward term $\left(N_{u}\right)$ is no longer necessary.

## Reference error format

In some cases the controller has access only to the signal, $e(k)=r(k)-y(k)$.


Example: Room thermostats for temperature control.

## Consequences?

State feedback/estimator design methods use separate measurements of $y(k)$ and $r(k)$.
Can we still do a state feedback design if we measure only $e(k)$ ?

## Estimation with only $e(k)$

Given a plant,

$$
\begin{aligned}
x(k+1) & =A x(k)+B u(k) \\
y(k) & =C x(k),
\end{aligned}
$$

we build an estimator and state feedback controller via,

$$
\begin{aligned}
\hat{x}(k+1) & =(A-B K) \hat{x}(k)+L(y(k)-C \hat{x}(k)) \\
u(k) & =-K x(k)
\end{aligned}
$$

or, equivalently,

$$
\begin{aligned}
\hat{x}(k+1) & =(A-B K-L C) \hat{x}(k)+L y(k) \\
u(k) & =-K x(k) .
\end{aligned}
$$

This isn't quite in the correct form. We cannot use $N_{x}$ or $N_{u}$ as $r(k)$ isn't available.

Estimation with only $e(k)$

$$
\begin{aligned}
\hat{x}(k+1) & =(A-B K-L C) \hat{x}(k)+L y(k) \\
u(k) & =-K x(k) .
\end{aligned}
$$

If we add an extra term, $-\operatorname{Lr}(k)$, we get,

$$
\begin{aligned}
\hat{x}(k+1) & =(A-B K-L C) \hat{x}(k)+L y(k)-\underbrace{L r(k)}_{\text {extra term }} \\
u(k) & =-K x(k) \quad
\end{aligned}
$$

This is now in a form we can implement with only an $e(k)$ measurement.

$$
\begin{aligned}
\hat{x}(k+1) & =(A-B K-L C) \hat{x}(k)-L e(k) \\
u(k) & =-K x(k) .
\end{aligned}
$$

The term $-L r(k)$ is an unwanted input to the estimator and will cause an offset in the estimation. If $r(k)=0$ this isn't a problem.

