

Lecture 3: Photodetectors

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Photodetectors (Continued)

⇒ Last lecture we covered

- ⇒ photoconductors and the dominant photon absorption mechanism: Intrinsic (Band-to-Band)
- ⇒ the power absorbed as a function of wavelength for different materials and derived the efficiency $\eta(\lambda, x)$
- ⇒ the concept of carrier lifetime and transit time and the resulting photoconductive gain (G) that results from a mismatch in electron and hole mobilities
- ⇒ Frequency response of photoconductive photodetectors

Biased p-n Junction Photodiodes



p-n Junction Photodiode Equation



 $I = (I_s)[\exp^{qV_{bias}/K_BT} - 1] - I_{photo}$ $= I_{dark} - I_{photo}$

• I_{dark} = is the current that occurs with zero optical input

 $\bullet I_s = I_{th}$ is the thermal or saturation current that occurs in normal (non-illuminated) diode operating mode

- I_{photo} is photo-generated current = $\frac{\eta q}{hv} P_{rcvd}$
- q is the electron charge
- V_{bias} is applied bias voltage (positive = forward, negative=reverse)
- K_B is Boltzman's constant
- T is temperature (usually in Kelvin, depending on units of K_B)

p-n Junction Photodiode Regions of Operation



pn-Junction Carrier Dynamics (1)

- ⇒ Carrier diffusion time (~ns/µm) is typically much longer than carrier transit time (~10ps/µm)
- ⇒ Electron and hole velocities saturated in depletion region due to high field strength
- ⇒ Once away from depletion region carrier velocities fall below saturation
- ⇒ Space charge barrier prevents carriers from entering the depletion region, therefore the multiple carrier effect seen in photoconductors does not occur when carrier velocities are mismatched



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pn-Junction Carrier Dynamics (2)

- Photons absorbed within one diffusion length outside the depletion region will be absorbed and the current contributing carriers will suffer both diffusion time and transit time delays
- ⇒ Effect is geometry and material dependent



pn-Junction Carrier Dynamics (3)

⇒ The separation of charge in the depletion region (due to uncompensated Donors and Acceptors) leads to a capacitive effect that also impacts the detector bandwidth





 $\varepsilon_0 = 8.85 \text{ x } 10^{-12} \text{ F/m} = \text{vacuum permitivitty}$ $\varepsilon_r = \text{semiconductor relative permitivitty}$ A = area of depletion region $l_d = \text{depletion region length}$

⇒ The frequency at which the detector bandwidth rolls off by 3-dB due to the junction capacitance is

$$\omega_{RC} \frac{1}{R_S C_j}$$

Cj = area of depletion region l_d = depletion region length

p-i-n Photodiodes

- To increase the photon absorption region, a layer of *intrinsic* semiconductor material can be added beteween the p and n material.
- The pin photodetector gain-bandwidth product improves of the pn-junction
 - ⇒ The detector quantum efficiency can be increased over that of a simple pn junction since the depletion region is almost entirely contained in the intrinsic region and the intrinsic region can be made long.
 - ⇒ Carrier diffusion effects minimized since all light absorbed in intrinsic region
 - The junction capacitance is reduced compared to a pn-junction because the distance between the effective plates is increased.
 - ⇒ Carriers reach saturation velocity while traveling in intrinsic region, so even though pin depletion length $l_p + l_d$ is longer than pn-junction depletion length, lower transit time than pn-junction where carrier velocity drops below saturation not far from metallurgical junction



p-i-n Photodiodes

⇒ As with the pn-junction, the quantum efficiency is defined by the following equation, however the distance can now be integrated over the larger intrinsic region

$$P_{abs}(x) = P_i(1-R)(1-e^{-\alpha(\lambda)x})$$
$$= \eta(\lambda, x)P_i$$

As the depletion region length is increased, η increases, the junction capacitance C_j decreases, and the transit time τ_{trans} increases. The detector design must be optimized to maximize both efficiency and bandwidth. An estimate of the bandwidth is given by

$$B_{pin} = \frac{1}{\sqrt{\left(\frac{1}{f_{RC}}\right)^2 + \left(\frac{1}{f_{trans}}\right)^2}} = \frac{1}{\sqrt{\left(2\pi R_S \varepsilon_0 \varepsilon_r \frac{A}{l_d}\right)^2 + \left(\frac{1}{0.44\nu_s}\right)^2}}$$

Bandwidth-Efficiency Tradeoffs in p-i-n Photodiodes

- ⇒ The quantum efficiency, η , can be approximated assuming R=0 (high quality anti-reflection coating) and intrinsic region length l_d . $\eta = 1 - e^{-\alpha l_d}$
- ⇒ For small l, bandwidth is transit time limited
- \Rightarrow For large l, bandwidth it RC limited
- ⇒ Optimal bandwidth length where two effects are equal
- ⇒ QE keeps increasing with increased length

 $\Rightarrow If the detector area A and length <math>l_d$ are both optimized, then bandwidth and quantum efficiency can both be maximized





Vertically Illuminated p-i-n Photodiodes



For a double pass vertically illuminated pin detector (see left figure), the quantum efficiency is

 $\eta = \left(1 + r e^{-\alpha l}\right) \left(1 - e^{-\alpha l}\right)$

When the carrier transit distance is approximately equal to l, and $\alpha l \ll 1$, the bandwidth-efficiency for a double-pass vertically illuminated pin photodiode is approximately

Uni-Traveling Carrier Photodiodes

Improving Saturation current using the high overshoot velocity of electrons to reduce the space charge effect – UTC

Photons are absorbed in the p-layer next to a wide bandgap undopedd drift layer

Holes are not transported -> become majority carriers very quickly

Electrons diffuse to depletion layer and drift rapidly

Thin depletion layer leads to electron overshoot drift velocities up to 5 times greater than saturation velocity

Space charge is reduced by factor of up to 5 relative to PIN, hence UTC

Avalanche Photodiodes (APDs)

 $\Rightarrow \alpha$ Rate at which electrons multiply

 $\Rightarrow \beta$ Rate at which holes multiply

- A large ratio of α/β or β/α results in a large gain bandwidth product and low noise amplification. True for Si
- ⇒ Most III-Vs have a small ratio, and limited gain bandwidth product. The noise is larger, but still lower than a PIN receiver.



InP Multiplication Layer

Ionization Coefficients for Semiconductors



SAM APDS:

Need for Separate Absorption and Multiplication Regions Small bandgap avalanche regions tend to have



Cross Section and Top View of SHIP Detector







Figure 5-4.3 The dc gain M of the avalanche photodiode (for electron injection at x=0) as a function of the number of multiplications α L and with the ionization ratio α/β as a parameter

Staircase APD: Use of bandgap engineering to increase the ratio of ionization coefficients.



no bias (above) and with reverse bias (below); (b) superlattice APD