

HW#5: Noise and Detection

1. (Device noise) So many devices or parts of devices in RF electronics display the canonical junction rectification I-V curve, $I = I_0 [\exp(eV/k_B T) - 1]$. Suppose the device is a one-port that displays generalized Johnson-Nyquist noise and full shot noise, and operates at 300 K into a load resistance R_L .
 - (a) For $R_L = 0$ (short-circuit case) and zero bias, what is the RMS current fluctuation through the device in terms of I_0 ? Interpret the answer in terms of the necessary device physics behind I_0 .
 - (b) If $R_L = 0$ (short-circuit case), at what bias voltage does the RMS shot-noise current equal the RMS thermal noise current?
 - (c) Suppose this device is connected to a standard $R_L = 50\text{-}\Omega$ load. At what bias voltage does the noise power generated in the load from the rectifier equal the noise power from the load itself?
 - (d) Express the noise power in (c) as a power spectral density of current fluctuations.

2. (System noise figure) Every linear component in a receiver chain can be represented by a noise figure, $F_i = (\text{SNR}_{\text{in}}/\text{SNR}_{\text{out}})_i = 1 + (N_{\text{add}})_i / (G_i k_B T_{300} B_i)$, where $(N_{\text{add}})_i$ is the noise added by the i th component, G_i is its gain, and B_i is its RF bandwidth.
 - (a) Use these definitions to derive Friis' formula for total noise figure, F_T .
 - (b) Suppose a receiver chain consists of a bandpass filter (BPF) having $G = -3.0$ dB and $F = 0.5$ dB, an LNA having $G = Y$ dB and $F = 2.0$ dB, a mixer having $G = -5$ dB and $\text{NF} = 6$ dB, and a power amp having $G = 30$ dB and $F = 10$ dB. What is the total noise figure if $Y = 30$ dB? [clue: easy and fun with an Excel spreadsheet]
 - (c) In the limit of very large Y , the noise figure approaches asymptotically a "front-end" limited value. What is this value? And to what value can Y be reduced so that the actual F is within 1 dB of the "front-end" limit?
 - (d) Suppose the BPF and LNA are interchanged. What is the new F_T if $Y = 30$ dB?

3. (Power-Law Detection) A common detector type used in RF sensors, particularly at low signal levels, is the square-law device represented by $X_{\text{out}}(t) = A[X_{\text{in}}(t)]^2 \equiv \mathfrak{R}P_{\text{in}}$.
 - (a) Of all the possible power-law detectors $X_{\text{out}} = A(X_{\text{in}})^n$, show that the square-law is the most effective at maximizing the ratio of baseband (dc) power to RF (ac) power.
 - (b) Assume a nonlinear diode is biased into a region where the I-V curve can be approximated to have the form $I = BV^2$. Use Taylor's theorem to write a general expression for the rectified current $\delta I = I - I_0$ in terms of the voltage difference δV induced at some bias point V_0 . Re-express this result as $\delta I = \mathfrak{R} P$, where $P = (\delta V)^2/R$ and \mathfrak{R} is the short-circuit responsivity. Write an expression for \mathfrak{R} in terms of known derivatives.
 - (c) One of the most useful devices through the history of THz sensors has been the metal-semiconductor, or Schottky, diode. A convenient model I-V curve for Schottky diodes is $I = I_S \{\exp(eV/nk_B T) - 1\}$, where n ("ideality" factor) > 1 . Formulate the square-law short-circuit \mathfrak{R} for a Schottky diode. Evaluate this responsivity at 300 K assuming $n = 1.0$ – the best possible case.