

ECE 137 A Mid-Term Exam

Thursday February 6, 2014

Do not open exam until instructed to.

Closed book: Crib sheet and 1 page personal notes permitted

There are 2 problems on this exam, and you have 75 minutes.

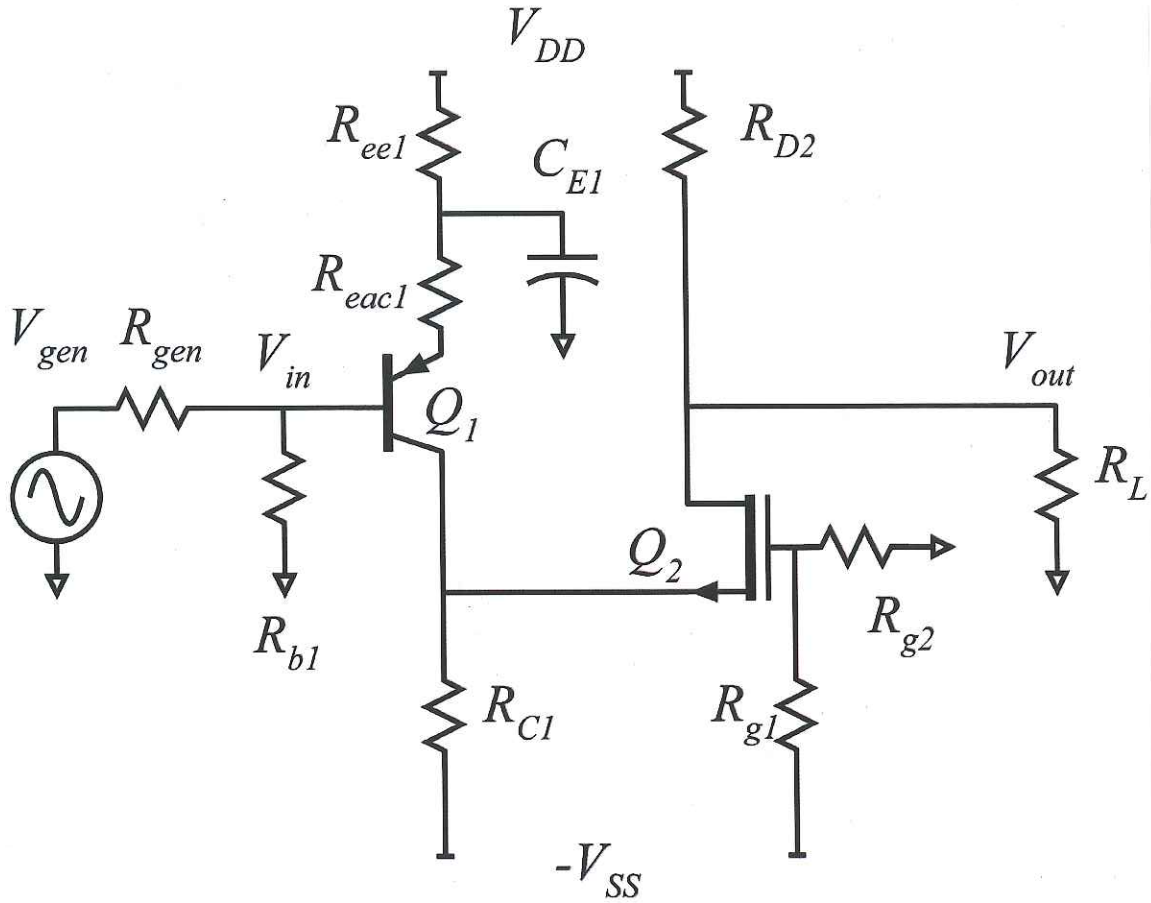
Use any and all reasonable approximations (5% accuracy is fine.), ***AFTER STATING and approximately Justifying them.***

Name: Solution "A"

Part	Points Received	Points Possible
1a		7
1b		7
1c		6
1d		15
1e		15
1f		6
1g		14
2a		12
2b		13
2c		5
TOTAL		100

Problem 1, 70 points

You will be working on the circuit below:



Q1: $\beta = 100$, $V_A = 100$ V

Q2: Velocity-limited $V_{th} = 0.2V$, $1/\lambda = \text{infinity}$, $\Delta V = L_g v_{th} / \mu = 0.1V$, $c_{ox} v_{th} W_g = 5\text{mA/V}$

The supplies are +3V and -3 V

$R_{gen} = 1000$ Ohms, $R_L = 10,000$ Ohms. $R_{g2} = 100\text{kOhms}$, $R_{eac1} = 74$ Ohms, $R_{b1} = 10\text{kOhms}$

C_{e1} is very large (AC short-circuit)

Part a, 7 points

DC bias.

V_{in} is at (approximately) zero volts DC.

The gate of Q2 is to be biased at -1.5 Volts

The drain is to be biased at zero volts.

Q1 is to be biased at 1 mA emitter current

Q2 is to be biased at 2 mA drain current.

Find the following:

$R_{c1} = 267\Omega$ $R_{g1} = 100k\Omega$ $R_{e1} = 2.226k\Omega$

$R_{g1} = 100k\Omega$

①
$$I_{D2} = 2mA = (5mA/V)(V_{gs1} - V_{th} - \Delta V)$$
$$2mA = (5mA/V)(V_{gs1} - 0.2V - 0.1V)$$
$$V_{gs1} = 0.2V + 0.1V + 2mA/(5mA/V) = 0.7V$$

①
$$V_{s1} = -0.7V - 1.5V = -2.2V$$

①
$$R_{c1} \text{ carries } 3mA \text{ (} I_{c1} + I_{D2} \text{)}$$
$$R_{c1} = \frac{3V - 2.2V}{3mA} = 267\Omega$$

① BASE of Q1 @ $0V \rightarrow V_{B1} \approx 0.7V$

①
$$R_{E1} = \frac{3V - 0.7V}{1mA} - R_{EAC1} = 2.3k\Omega - 74\Omega = 2.226k\Omega$$

①
$$R_{D2} = \frac{3V - 0V}{2mA} = 1.5k\Omega$$

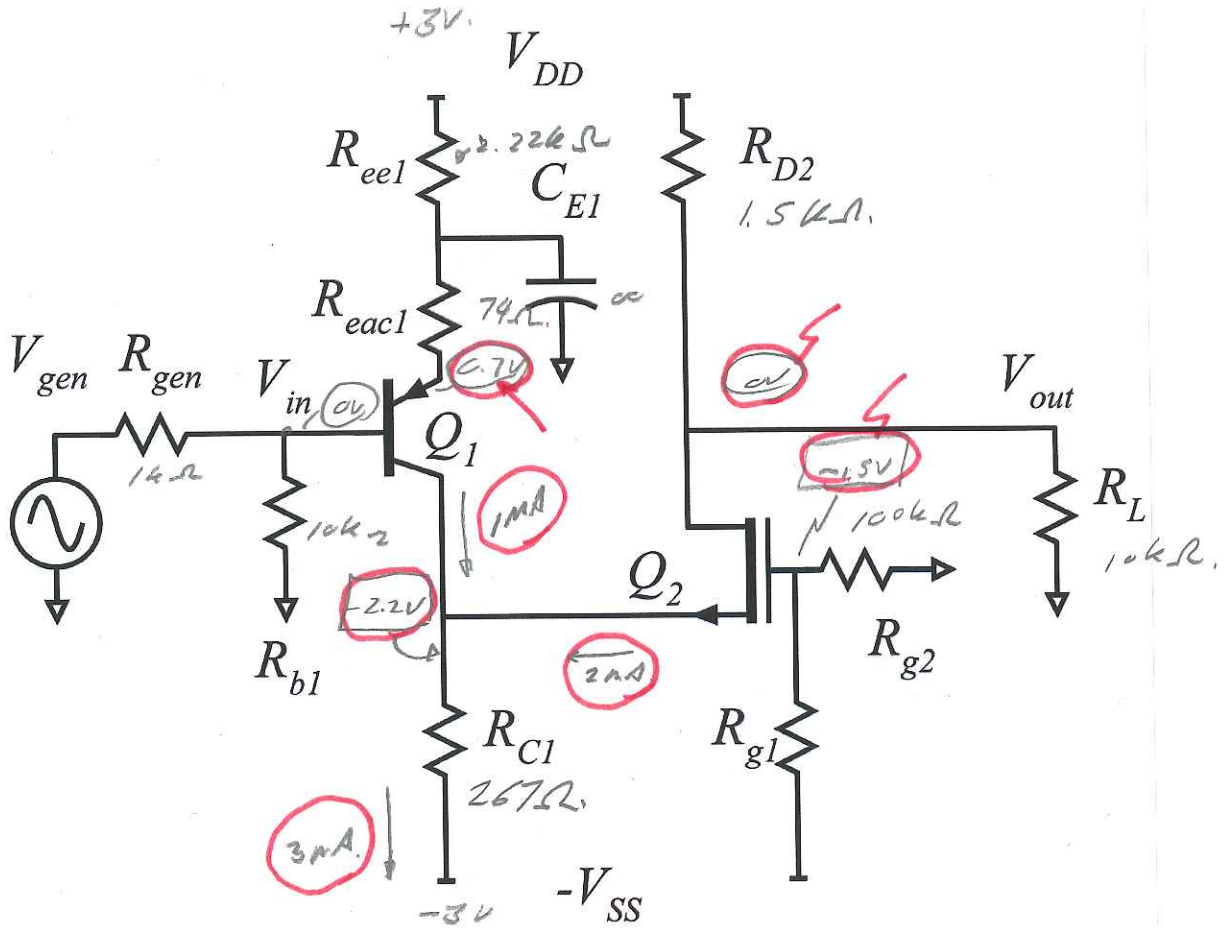
①
$$R_{g2} \text{ carries } 1.5V/100k\Omega = 15\mu A$$
$$R_{g1} = \frac{3V - 1.5V}{15\mu A} = 100k\Omega$$

$\alpha_2 : V_{th} = 0.2V, I = 0, \Delta V = 0.1V$
 $5 \mu A/V$

$Q_1 : \beta = 100, V_A = 100V$

Part b, 7 points

DC bias



On the circuit diagram above, label the DC voltages at ALL nodes and the DC currents through ALL resistors

Part c, 6 points

Find the small signal parameters of Q1 and Q2.

Transistor Q1: $g_m = 38.5 \text{ mS}$ $R_{ce} = 100 \text{ k}\Omega$ $R_{be} = 2.6 \text{ k}\Omega$

Transistor Q2: $g_m = 5 \text{ mS}$ $R_{ds} = \infty \Omega$

Q1: $I_C = 1 \text{ mA}$, $V_A = 100 \text{ V}$, $\beta = 100$

1.5 $\left[g_m = \frac{1 \text{ mA}}{26 \text{ mV}} = 38.5 \text{ mS} = \frac{1}{26 \Omega} \right]$

1.5 $\left[R_{be} = \beta / g_m = 100 \cdot 26 \Omega = 2.6 \text{ k}\Omega \right]$

1 $\left[R_{ce} = \frac{V_{CE} + V_A}{I_C} \approx \frac{V_A}{I_C} = \frac{100 \text{ V}}{1 \text{ mA}} = 100 \text{ k}\Omega \right]$

Q2: $I_D = 2 \text{ mA}$

1 $\left[g_m = 5 \text{ mS/V} \right]$

1 $\left[R_{ds} = \infty \right]$ because $\lambda = 0 \text{ V}^{-1}$ velocity limited

Part d, 15 points.

Find the small signal voltage gain (V_{d2}/V_{s2}) of Q2 and Q2's small-signal input resistance.

$$V_{d2}/V_{s2} = \underline{\quad + 6.5 \quad}$$

$$R_{in,q2} = \underline{\quad 200 \Omega \quad}$$

3.5 [Q2: Common-gate.

3.5 [BECAUSE $r_{DS2} = \infty$, $R_{in2} = 1/g_{m2} = \frac{1}{5mS} = 200 \Omega$.

3.5 [$R_{Leg2} = R_{D2} \parallel R_L = 1.5k\Omega \parallel 10k\Omega = 1.3k\Omega$

3.5 [$\frac{V_{d2}}{V_{s2}} = \frac{R_{Leg2}}{R_{in2}} = \frac{1.3k\Omega}{200\Omega} = 6.5$

Part e, 15 points

Find the small signal voltage gain (V_{c1}/V_{b1}) of Q1 and the *** amplifier *** input resistance.

$V_{c1}/V_{b1} = \underline{-1.14}$

$R_{in, amplifier} = \underline{5k\Omega}$

2 [Q1 \rightarrow common-emitter with degeneration.

$1/g_m = 26\Omega, R_{ECC} = 74\Omega$

2 [$R_{out, c1} = R_C (1 + g_m R_{ECC}) = 100k\Omega (1 + 74/26\Omega)$
 \rightarrow very large, can neglect.

3 [$R_{L, eq1} = R_{C1} \parallel R_{in2} \parallel R_{out, c1}$
 $\approx 267\Omega \parallel 200\Omega = 114\Omega$

3 [$A_{v1} = \frac{-R_{L, eq1}}{r_{ECC} + 1/g_m} = \frac{-114\Omega}{26\Omega + 74\Omega} = -1.14$

3 [$R_{in, g1} = \beta (1/g_m + R_{ECC1}) = 100 \cdot (74\Omega + 26\Omega)$
 $= 10,000\Omega$

2 [$R_{in, \Omega} = R_B \parallel R_{in, g1} = 10k\Omega \parallel 10k\Omega = 5k\Omega$

Part f, 6 points

Find (V_{out}/V_{in}) , (V_{in}/V_{gen}) and (V_{out}/V_{gen})

$$(V_{out}/V_{in}) = \underline{\underline{-7.41}}$$

$$(V_{in}/V_{gen}) = \underline{\underline{0.833}}$$

$$(V_{out}/V_{gen}) = \underline{\underline{-6.17}}$$

$$2 \left[\begin{aligned} V_{in}/V_{gen} &= \frac{R_{inA}}{R_{inA} + R_{gen}} = \frac{5k\Omega}{5k\Omega + 1k\Omega} \\ &= 0.833 \end{aligned} \right.$$

$$2 \left[\begin{aligned} V_{o1}/V_{in} &= A_{v1} A_{v2} = -1.14 \cdot 6.5 = -7.41 \end{aligned} \right.$$

$$2 \left[\begin{aligned} V_{o1}/V_{gen} &= V_{o1}/V_{in} \cdot V_{in}/V_{gen} \\ &= -7.41 \cdot 0.833 \\ &= -6.17 \end{aligned} \right.$$

Part g, 14 points

Now you must find the maximum signal swings. Find the output voltage due to saturation and cutoff in Q2. **Give the sign (+ or -) in your answers below.**

Cutoff of Q1; Maximum ΔV_{out} resulting = -0.74V.

Saturation of Q1; Maximum ΔV_{out} resulting = +15.6V

Cutoff of Q2; Maximum ΔV_{out} resulting = +2.6V.

Knee voltage of Q2; Maximum ΔV_{out} resulting = -2.1V.

Note: $A_{v2} = +6.5$

3.5 Cutoff Q1:

- 1. $\Delta I_{C1max} = 1mA - 0mA = 1mA$ (↓) decrease
- 1. $R_{Leg1} = 114\Omega$
- 1. $\Delta V_{C1max} = 1mA \cdot 114\Omega = 114mV$, negative-going ↓
- 1/2 $\Delta V_{out} = 114mV \cdot A_{v2} = 114mV \cdot (6.5) = 0.741V$ ↓

3.5 Saturation of Q1

- 1. $V_{CEQ} = 0.7V + 2.2V = 2.9V$, $V_{CEsat} \approx 0.5V$
- 1. $\Delta V_{CEmax} = 2.4V$; neglecting ΔV_E , $\Delta V_{C1max} = 2.4V$ ↑
- 1/2 $\Delta V_{out} = 2.4V \cdot (6.5) = +15.6V$

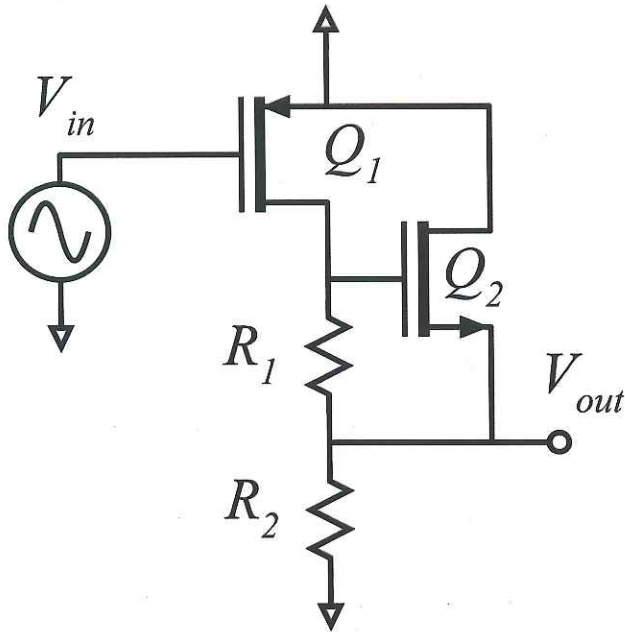
3.5 Cutoff Q2

- 1.5 $I_{DQ} = 2mA$, $\Rightarrow \Delta I_{Dmax} = 2mA$ (decrease)
- 2 $\Delta V_{out} = 2mA \cdot R_{Leg2} = 2mA \cdot 1.3k\Omega = 2.6V$

3.5 Knee Q2

- 1 $V_{DS0} = 2.2V$, $V_{DSknee} = 0.1V$ (velocity-saturated)
- 1/2 $\Delta V_{DS} \approx \Delta V_{out} = 2.1V$ ↓

Problem 2, 30 points
nodal analysis



You will be working on the circuit to the left.

Ignore DC bias analysis. You don't need it.

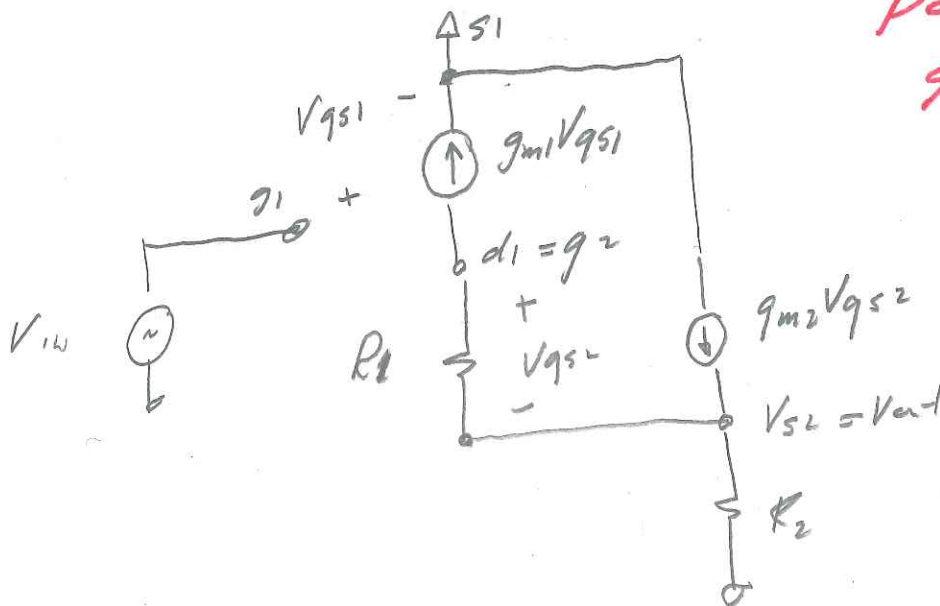
Transistor 1 has transconductance g_{m1} .

Transistor 2 has transconductance g_{m2} .

The drain-source resistances R_{ds} of both transistors are infinity (so you don't need to draw it!)

Part a, 12 points

Draw the small-signal equivalent circuit



*points off if
gm generators are
not labelled
or controlling voltages
not identified*

Part b, 13 points

Find, by nodal analysis, a small-signal expression for V_{out}/V_{in} .

$V_{out}/V_{in} =$ _____

there are two unknowns: $V_{d1} = V_{g2}$ and V_{out} .

$\Sigma I = 0$ @ V_{g2} (+ = leaving)

$$4 \left[+g_{m1} V_{gs1} + (V_{g2} - V_{out}) / R_1 = 0 \right]$$
$$+ g_{m1} (V_{in}) + V_{g2} / R_2 - V_{out} / R_2 = 0$$

$$V_{g2} (1/R_2) + V_{out} (-1/R_2) = -g_{m1} V_{in}$$

$\Sigma I = 0$ @ V_{out} (+ = leaving)

$$4 \left[(V_{out} - V_{g2}) / R_1 + V_{out} / R_2 - g_{m2} V_{gs2} = 0 \right]$$
$$(V_{out} - V_{g2}) / R_1 + V_{out} / R_2 - g_{m2} (V_{g2} - V_{out}) = 0$$

$$V_{g2} (-1/R_1 - g_{m2}) + V_{out} (1/R_1 + 1/R_2 + g_{m2}) = 0$$

$$\begin{bmatrix} 1/R_1 & -1/R_1 \\ -1/R_1 - g_{m2} & 1/R_1 + 1/R_2 + g_{m2} \end{bmatrix} \begin{bmatrix} V_{g2} \\ V_{out} \end{bmatrix} = \begin{bmatrix} -g_{m1} \\ 0 \end{bmatrix} V_{in}$$

So $\frac{V_{out}}{V_{in}} = \frac{N}{D}$ where

$$D = \begin{vmatrix} 1/R_1 & -1/R_1 \\ -1/R_1 - g_{m2} & 1/R_1 + 1/R_2 + g_{m2} \end{vmatrix}$$

$$= (1/R_1)(1/R_1 + 1/R_2 + g_{m2}) - (1/R_1)(1/R_1 + g_{m2})$$

$$N = \begin{vmatrix} 1/R_1 & -g_{m1} \\ -1/R_1 - g_{m2} & 0 \end{vmatrix} = -g_{m1}(1/R_1 + g_{m2})$$

$$\frac{V_{out}}{V_{in}} = \frac{-g_{m1}(1/R_1 + g_{m2}) R_1}{(1/R_1 + 1/R_2 + g_{m2}) - (1/R_1 + g_{m2})}$$

$$\frac{V_{out}}{V_{in}} = -g_{m1}(1 + g_{m2} R_1) R_2 \leftarrow S.$$

Part c, 5 points

$g_{m1} = 1 \text{ mS}$ $g_{m2} = 10 \text{ mS}$, $R_1 = 1 \text{ k}\Omega$, $R_2 = 2 \text{ k}\Omega$
Give a numerical value for V_{out}/V_{in} .

$V_{out}/V_{in} = \underline{\underline{-22}}$

5

$$\left[\begin{aligned} \frac{V_{out}}{V_{in}} &= -1 \text{ mS} (1 + 10 \text{ mS} \cdot 1 \text{ k}\Omega) \cdot 2 \text{ k}\Omega \\ &= -1 \text{ mS} (11) \cdot 2 \text{ k}\Omega \\ &= -11 \text{ mS} \cdot 2 \text{ k}\Omega \\ &= -22 \end{aligned} \right.$$