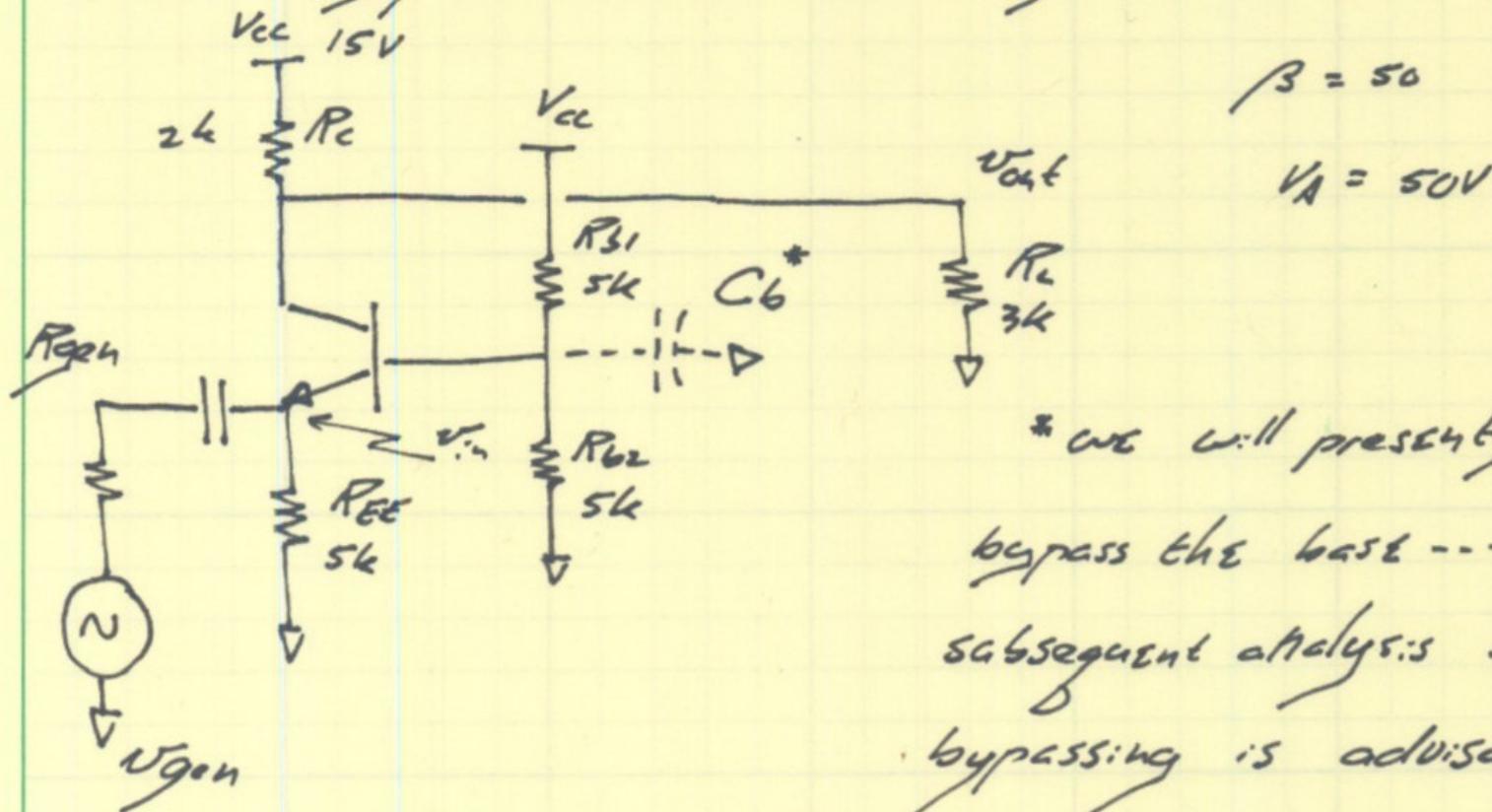


ECE137A, Notes Set 6: Common Gate, Common Base

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Common-base stage

First - analyze Common-base stage:

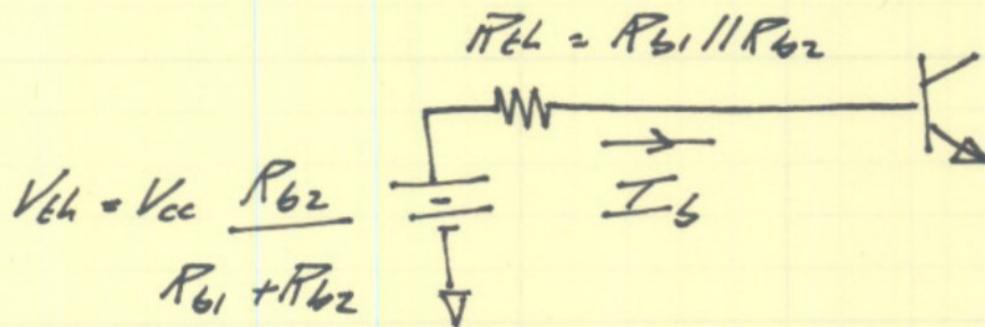


* we will presently not ac
bypass the base --- but
subsequent analysis shows that
bypassing is advisable.

Bias Analysis: Iterative

Assume $\beta \rightarrow \infty \Rightarrow I_B = 0 \Rightarrow V_{BE} = 7.5V \Rightarrow V_E = 6.8V$
 $\Rightarrow I_E = 6.8V / 5k\Omega = 1.36 \text{ mA}$

Now assume $\beta = 50 \Rightarrow I_B = I_E / \beta = 27 \mu\text{A}$



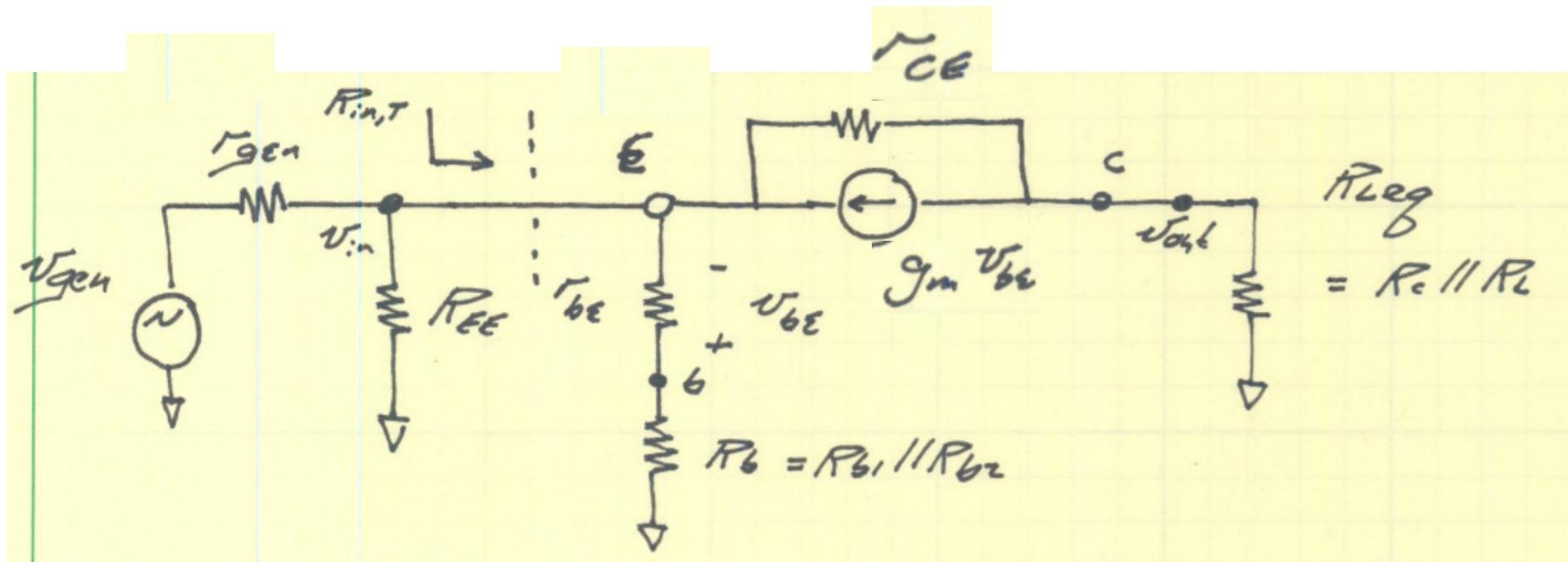
$$\Rightarrow V_E = 7.5V - I_B \cdot R_{BL} = 7.5V - 2.5k\Omega \cdot 27 \mu\text{A}$$

$$= 7.5V - 68mV$$

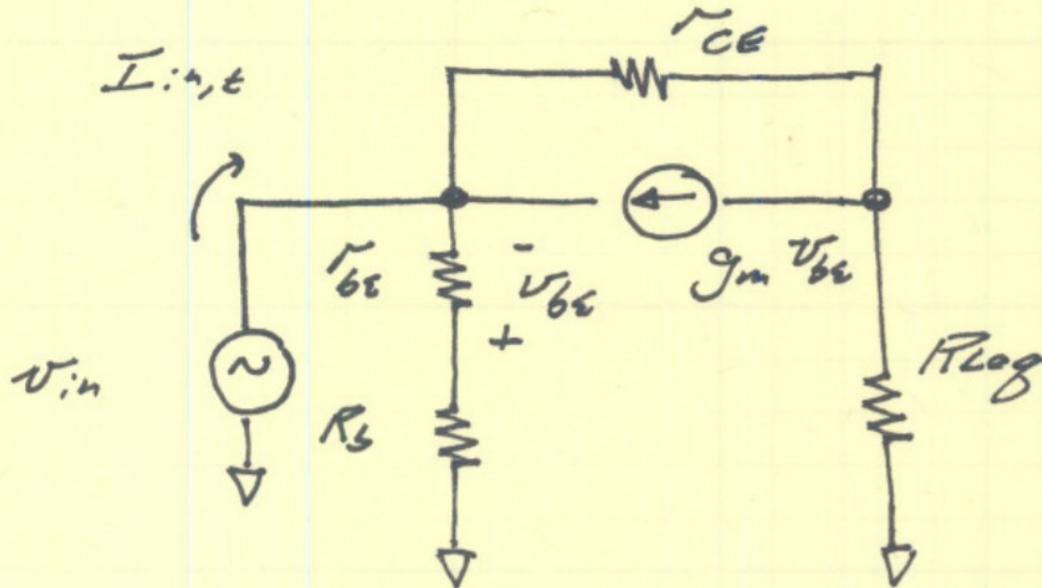
a negligible correction

→ need not further analyze for
corrections in I_E .

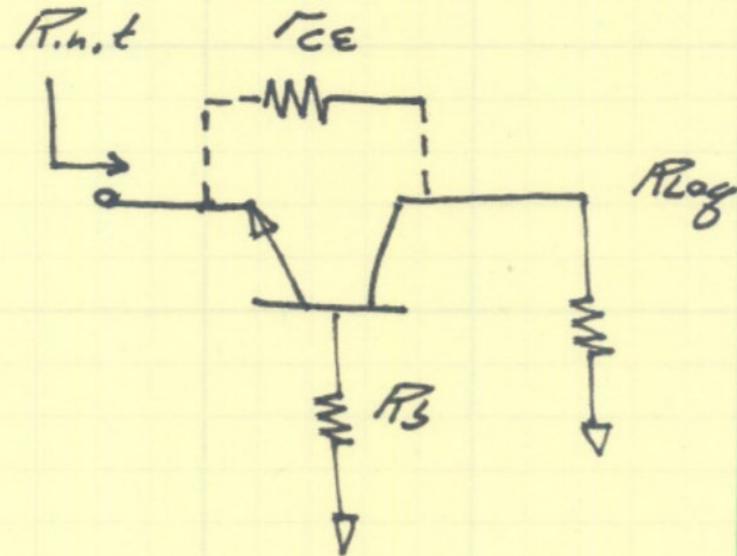
Nearly exact small-signal analysis



Finding transistor input impedance:

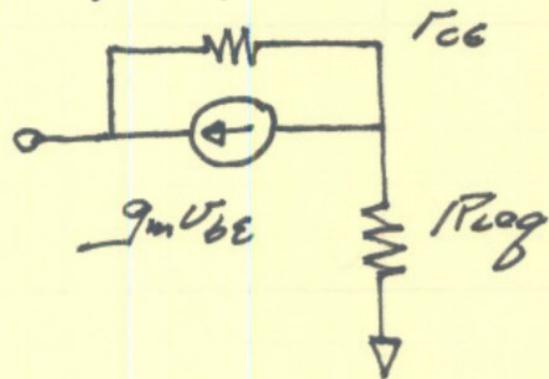


$$V_{in} / I_{in,t} \triangleq R_{in,t}$$

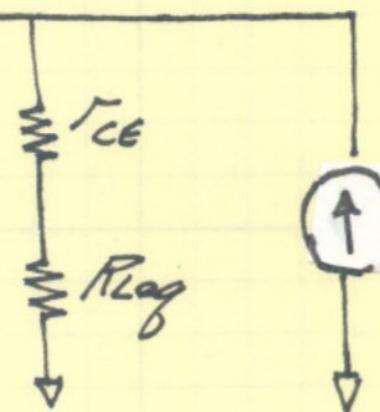


Finding transistor input impedance:

To simplify, and to understand,
recognize the current divider in the circuit:



Norton Eq.

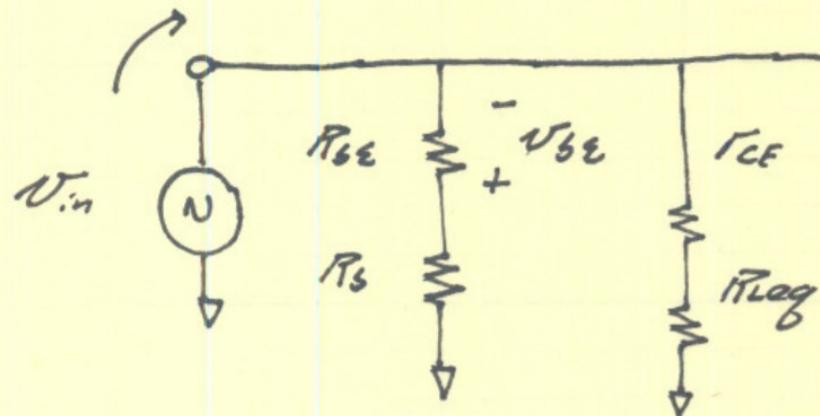


$$g_m V_{be} \cdot \frac{R_{ce}}{R_{ce} + R_{Leq}}$$

Finding transistor input impedance:

So we are trying to find R_{in} of the network below:

$I_{in,t}$



$$I = g_m V_{be} \cdot \frac{R_{ce}}{R_{ce} + R_{leq}}$$

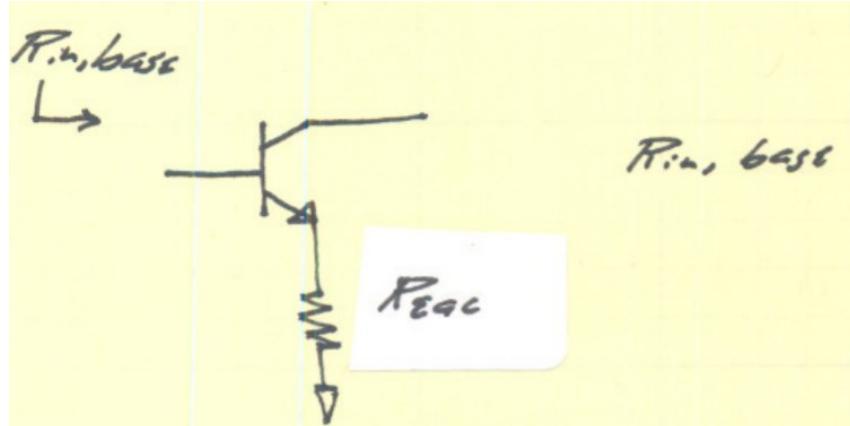
$$= -g_m V_{in} \cdot \frac{R_{be}}{R_{be} + R_b} \cdot \frac{R_{ce}}{R_{ce} + R_{leq}}$$

The current in the right-most branch is dominant, so

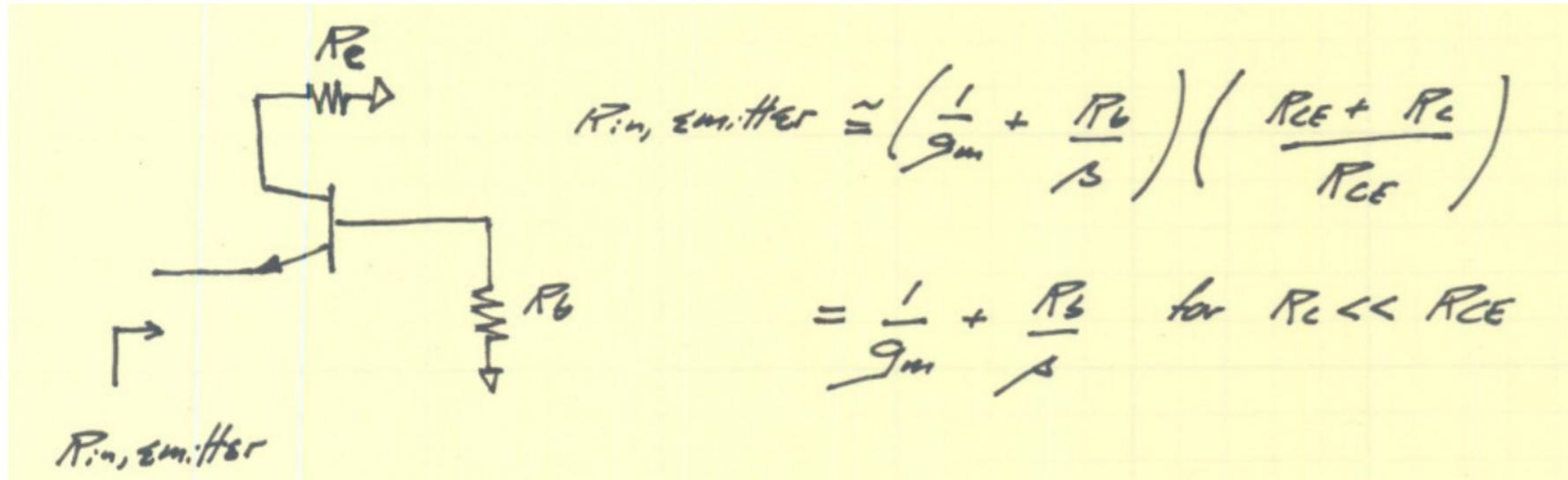
$$R_{in,t} \approx \frac{V_{in}}{I} = \frac{R_{be} + R_b}{g_m R_{be}} \quad \frac{R_{leq} + R_{ce}}{R_{ce}} = \left(r_e + \frac{R_b}{\beta} \right) \left(\frac{R_{leq} + R_{ce}}{R_{ce}} \right)$$

$$R_{in,t} \approx \left(r_e + \frac{R_b}{\beta} \right) \left(\frac{R_{leq} + R_{ce}}{R_{ce}} \right) = \left(\frac{1}{g_m} + \frac{R_b}{\beta} \right) \left(\frac{R_{leq} + R_{ce}}{R_{ce}} \right)$$

Some pictures:



$$R_{in, base} \approx \beta (r_E + R_{eac})$$

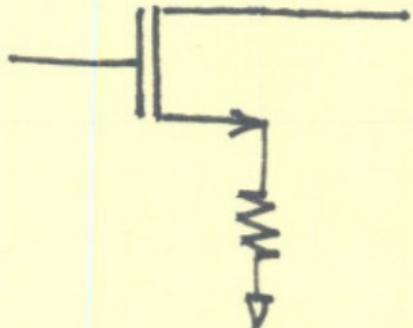


$$R_{in, emitter} \approx \left(\frac{1}{g_m} + \frac{R_E}{\beta} \right) \left(\frac{R_{CE} + R_C}{R_{CE}} \right)$$

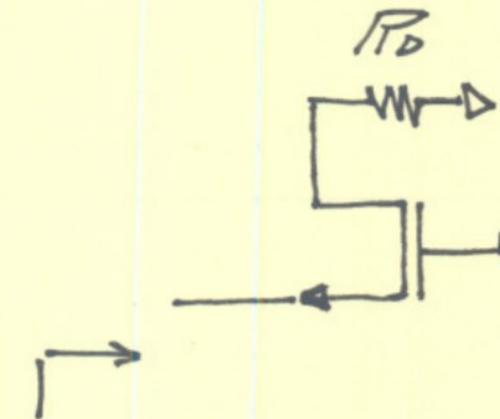
$$= \frac{1}{g_m} + \frac{R_E}{\beta} \quad \text{for } R_C \ll R_{CE}$$

Some pictures:

$R_{in, gate}$



$R_{in, gate} = \infty$, of course

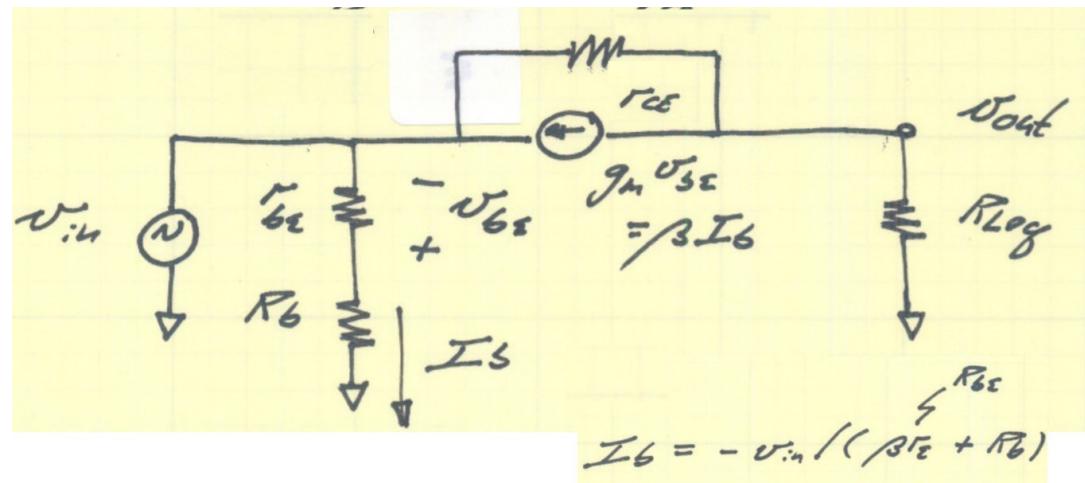
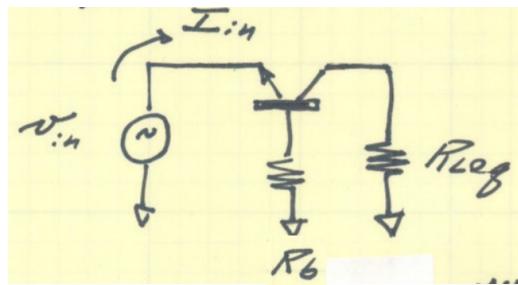


$R_{in, source}$

$$R_{in, source} = \frac{1}{g_m} \left(\frac{R_{ds} + R_D}{R_{ds}} \right)$$

$$= \frac{1}{g_m} \quad \text{for } R_D \ll R_{ds}$$

Common-base voltage gain



nodal equation at \$V_{out}\$:

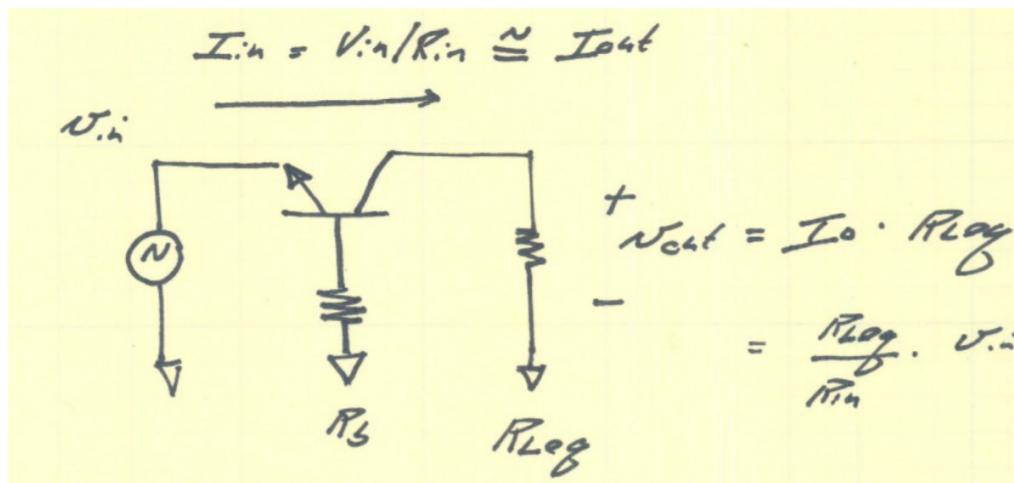
$$0 = V_{out} \left(\frac{1}{R_{load}} + \frac{1}{r_{ce}} \right) + V_{in} \left(-\frac{1}{r_{ce}} - \frac{\beta}{\beta r_{ce} + R_B} \right)$$

now \$r_{ce} \gg r_e + R_B/\beta\$ unless \$R_B\$ is really big...

$$\Rightarrow \frac{V_{out}}{V_{in}} = \frac{1}{r_e + R_B/\beta} \quad \frac{1}{1/R_{load} + 1/r_{ce}}$$

Common-base voltage gain

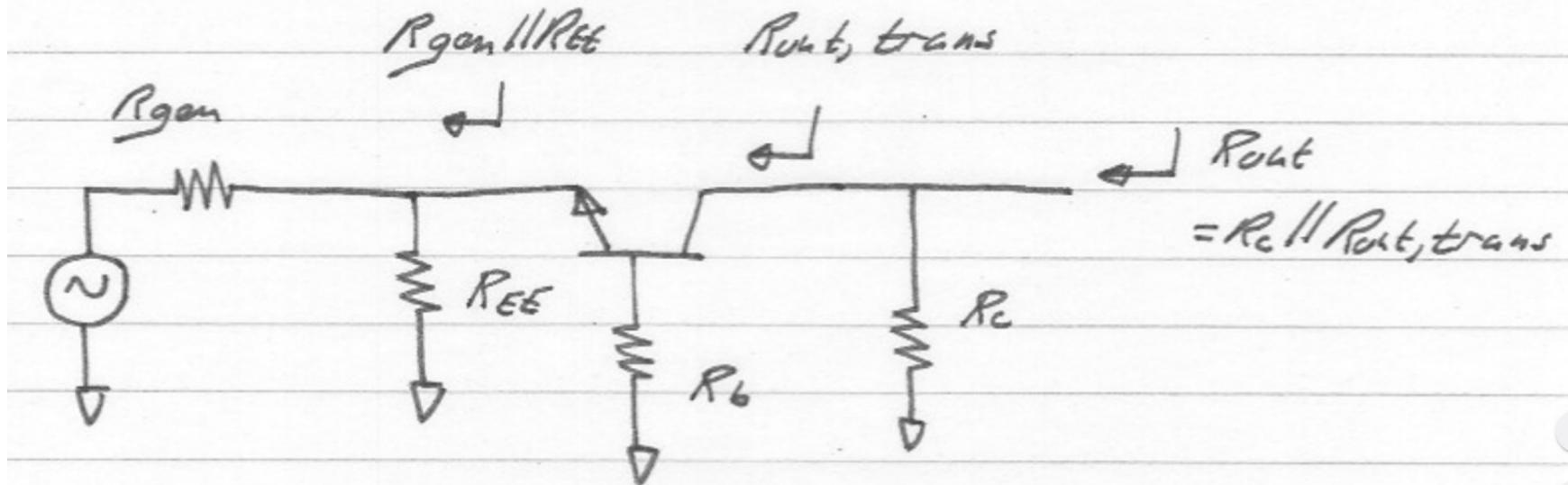
$$\frac{V_{out}}{V_{in}} = \frac{R_{Leq}}{r_e + R_b / \beta} \cdot \frac{R_{ce}}{R_{ce} + R_{Leq}} = \frac{R_{Leq}}{R_{in,emitter}}$$



For the FET, we have the same relationship:

$$\frac{V_{out}}{V_{in}} = \frac{R_{Leq}}{1/g_m} \cdot \frac{R_{DS}}{R_{DS} + R_{Leq}} = \frac{R_{Leq}}{R_{in,source}}$$

Output impedance: Common-Base



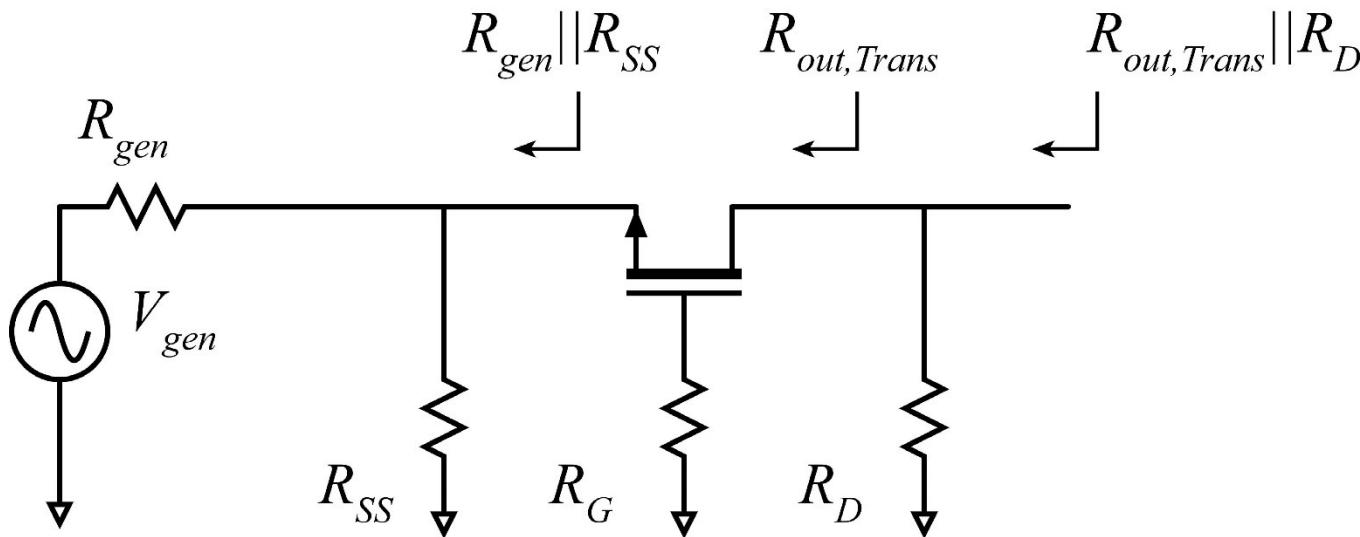
$$R_{out,Trans} = R_{CE} \left[1 + \frac{R_{gen} \parallel R_{EE}}{1/g_m} \frac{R_{be}}{R_{be} + R_b + R_{gen} \parallel R_{EE}} \right] \approx \left[\frac{R_{gen} \parallel R_{EE}}{1/g_m} \right]$$

this is the same analysis as the common-emitter + degeneration output impedance

For the FET this becomes

$$R_{out,Trans} = R_{DS} \left[1 + \frac{R_{gen} \parallel R_{SS}}{1/g_m} \right]$$

Output impedance: Common-Gate



For the FET, the analysis is identical
(except for the absence of R_{be}) and hence:

$$R_{out,Trans} = R_{DS} \left[1 + \frac{R_{gen} \parallel R_{SS}}{1/g_m} \right]$$