

## 3

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# Frequency Response of Transistor Amplifiers

## Overview

This lab will continue our exploration of single-transistor amplifiers using BJTs. We will explore the frequency response of some simple BJT amplifiers, and examine the factors affecting the low- and high-frequency cutoff frequencies in these amplifiers.

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## Pre-lab Preparation

### Before Coming to the Lab

Read through the lab experiment to familiarize yourself with the components and assembly sequence. Before coming to the lab, each group should obtain a parts kit from the ECE Shop.

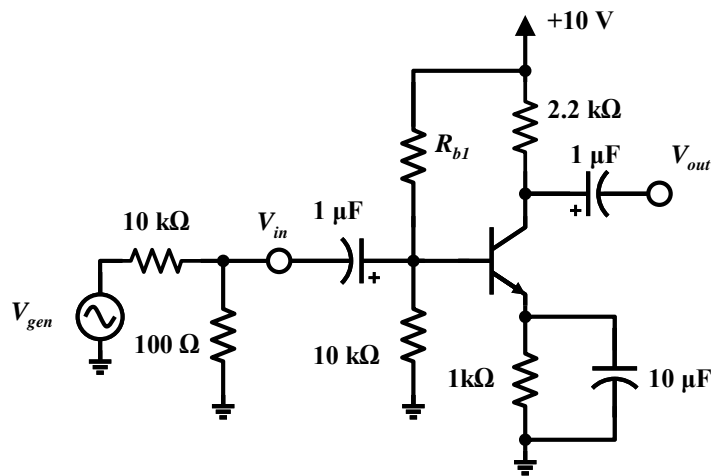
### Parts List

Use parts from Lab #2

## In-Lab Procedure

### 3.1 Common-Emitter Amplifier

Figure 3-1 shows an AC-coupled common-emitter amplifier, similar to the CS amplifiers from the previous lab. We will use this circuit to explore the frequency response of the CE/CS amplifier topology. Since the circuit has a rather high gain, we have included a resistor divider network at the input to keep the input signal small.



- Figure 3-1 – Common-emitter amplifier (with input divider).
- Using a 2N3904 in the circuit of Figure 3-1, first calculate the base resistance  $R_{b1}$  that is needed to bias the device at a collector voltage of  $V_c \approx 6 - 7\text{ V}$ . Assume a nominal value of  $\beta \approx 200$  for this calculation (document your work in your lab report).
  - Build the circuit using your calculated value of  $R_{b1}$ , and record the DC voltage at the collector, base, and emitter. From these measurements, estimate the collector current  $I_c$ , and transconductance  $g_m$ .

### Low-Frequency Response

- Adjust the function generator for a 0.1 V amplitude sinewave at 10 kHz and apply to the circuit using the voltage divider shown. Record the waveforms at the points  $V_{in}$  and  $V_{out}$ , and compute the gain of the amplifier circuit  $A_{vo} = V_{out}/V_{in}$ . This is the mid-band gain of the amplifier.

Note: it may be hard to see  $V_{in}$  directly on the scope since it is a small signal. If so, using signal averaging on the oscilloscope may help. Another alternative is to temporarily increase the input signal until  $V_{in}$  is easily measurable and the step-down ratio of the input voltage divider can be determined accurately (for example, in Figure 3-1 the input divider reduces the

input signal by a factor of 100 nominally). From that point on you can just observe  $V_{gen}$  on the scope and apply your measured scaling factor to indirectly determine  $V_{in}$ .

- From the measured gain, estimate the transconductance and compare with the value determined from DC measurements.
- Slowly decrease the frequency until the output signal reduces by -3dB ( $1/\sqrt{2}$  of its original value). This is the low-frequency cutoff,  $f_L$ .
- Reduce the frequency to  $f_L/10$  and measure the gain again. Repeat for  $f_L/100$ .
- The pole at  $f_L$  is due primarily to the RC time-constant associated with the emitter bypass capacitor. Verify this by replacing the  $10\mu\text{F}$  bypass capacitor by a  $100\mu\text{F}$  capacitor and repeating the last two steps.

In your lab report, compare your measurements of the low-frequency cutoff against the theoretical value.

### High-Frequency Response

Now, replace the  $10\mu\text{F}$  bypass capacitor and return the function generator to a 0.1V sinewave at 10kHz.

- Increase the frequency to find the high-frequency -3dB cutoff of the amplifier,  $f_H$ .

As we learned in class, the upper cutoff is determined by the RC time-constant formed by the internal capacitances of the transistor and the surrounding resistances. We can demonstrate this as follows.

- Simulate the effect of increasing  $C_{be}$  and  $C_{bc}$  by adding an external capacitance between the base and emitter. Add a  $0.01\mu\text{F}$  ceramic capacitor in this way, and re-measure the high-frequency cutoff.
- Also measure the gain for  $f = 10f_H$  in this case.

In your lab report, use the measured data to create a Bode plot of the gain-frequency response of the amplifier in Figure 3-1.

## 3.2 Common-Base Amplifier

The basic circuit is shown in Figure 3-2 and is quite similar to the common-gate circuit from the previous lab. Here we have simplified the topology to eliminate the AC input/output coupling capacitors, and included a voltage divider at the input to keep the drive signal small.

- First assemble the circuit using a collector resistor of  $R_c = 2.2\text{k}\Omega$ , as in the CE amplifier. Note the supply voltage is now +12V.
- Apply a 1V sinewave at 10kHz. Observe and record the waveforms at the  $V_{in}$  and  $V_{out}$  nodes (use AC coupling on the oscilloscope) and from this information calculate the mid-band gain.

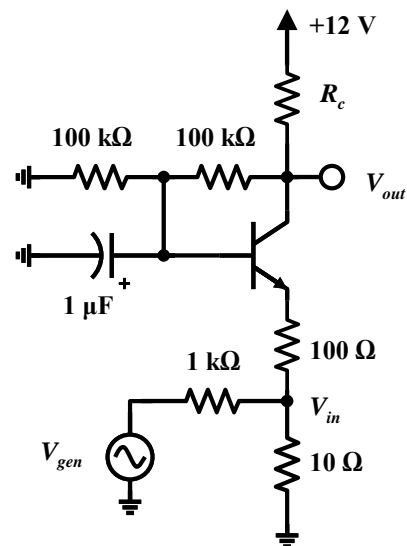


Figure 3-2 – CB amp.

Note that we could increase the gain by adding a bypass across the  $100\Omega$  emitter resistor, but we have left that out here to simplify the analysis.

### Low-Frequency Response

- Reduce the frequency to find the low-frequency cutoff  $f_L$ . Verify that this is due to the bypass capacitor at the base by increasing this capacitor and re-measuring the low-frequency cutoff. Record your results.

### High-Frequency Response

As discussed in the text and lecture, the common-base/common-gate topology has an inherently high cutoff frequency relative to the CE stage.

- Increase the frequency and try to find the high-frequency -3dB cutoff of the amplifier,  $f_H$ . If the function generator doesn't have the frequency range to find this frequency, state this in your report.
- Simulate the effect of increasing  $C_{be}$  and  $C_{bc}$  by adding an external capacitance between the base and emitter. Add a 0.01 $\mu$ F ceramic capacitor in this way, and re-measure the high-frequency cutoff if you can.

To see that the internal capacitances impose a limit on the gain-bandwidth product of the amplifier, let's increase the gain and observe the effect on the frequency response:

- Remove the external capacitance you added in the previous step and increase the gain by changing the collector resistor to  $R_c = 10\text{ k}\Omega$ .
- Now re-measure the mid-band gain at 10kHz and find new high-frequency cutoff.
- Repeat the last step with the 0.01 $\mu$ F external base-emitter capacitance.

In your lab report, use your measured data to sketch a Bode plot for the common-base amplifier, and compare with theoretical values for mid-band gain and cutoff frequencies.

## 3.3 Multi-Stage Amplifier Example

The circuit in Figure 3-3 is a cascade of two common-emitter stages, one with an NPN device and the second using a PNP device. Cascading gain stages has the advantage of increased gain, but comes at the expense of a decreased bandwidth, as you will see firsthand below.

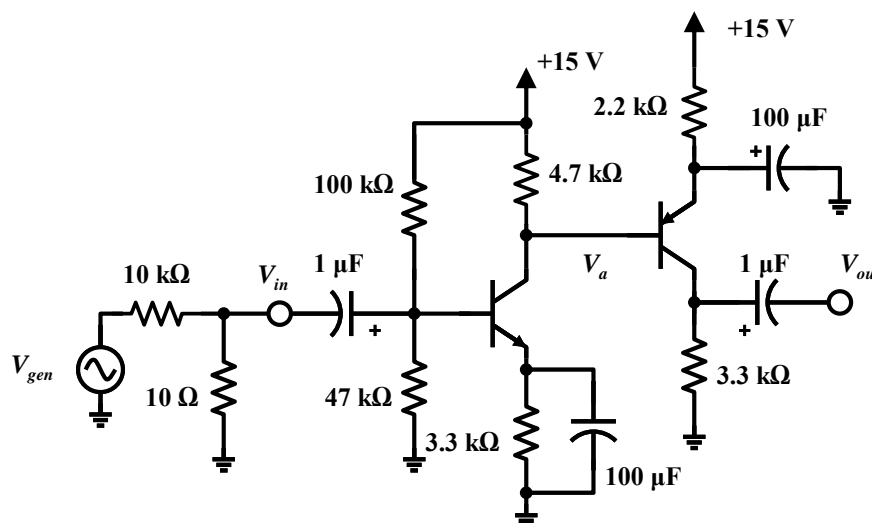


Figure 3-3 – Two-stage amplifier.

- Construct the circuit in circuit Figure 3-3. Note the supply voltage has been increased to 15 V. Measure and record the mid-band AC voltage gain  $A_{vo} = V_{out} / V_{in}$  with a generator signal of 0.1V sinusoid at 10kHz. It may be difficult to see  $V_{in}$  on your oscilloscope because of the large step-down ratio of the voltage divider at the input (nominally a divide-by-1000). If so, increase the input signal briefly to establish the actual voltage division ratio, then use this information to calculate the  $V_{in}$  for a 0.1V generator input.
- Decrease the frequency to find the low-frequency cutoff,  $f_L$ . Record this and then reduce the frequency to  $f_L / 10$  and measure the gain again. Repeat for  $f_L / 100$ . Can you explain your results?
- Now increase the frequency to find the high-frequency cutoff,  $f_H$ . Record this and then increase the frequency to  $10f_H$  and measure the gain again. Repeat for  $100f_H$  (if the function generator permits). Can you explain your results?

In your lab report, use your measured data to sketch a Bode plot for the gain-frequency response of this amplifier. Compare your results with the expected gain and cutoff frequencies.

Congratulations!  
You have now completed Lab 3