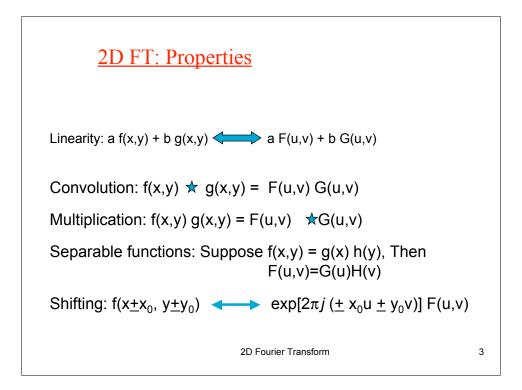


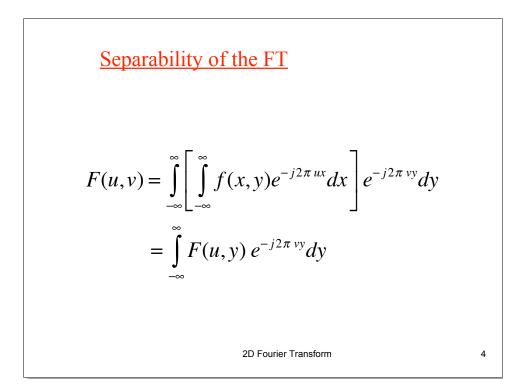
Fourier Transform - review  

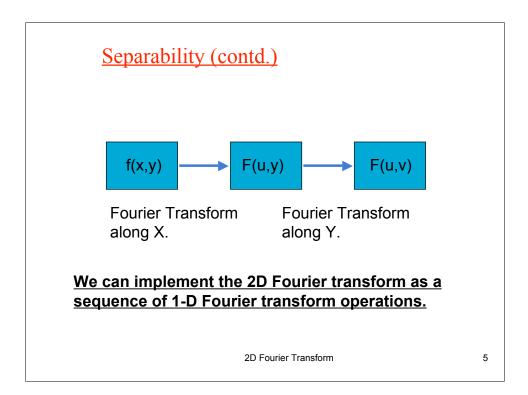
$$F(u) \equiv \Im\{f(x)\} = \int_{-\infty}^{\infty} f(x)e^{-j2\pi ux} dx$$
1-D:  

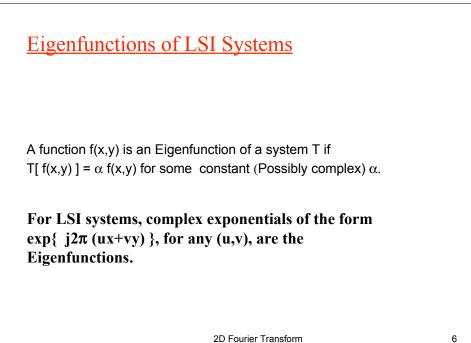
$$f(x) \equiv \Im^{-1}\{F(u)\} = \int_{-\infty}^{\infty} F(u)e^{j2\pi ux} du$$
F(u,v) =  $\iint f(x,y)e^{-j2\pi(ux+vy)} dx dy$   
2-D:  

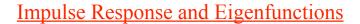
$$f(x,y) = \iint F(u,v)e^{j2\pi(ux+vy)} du dv$$
2D Fourier Transform





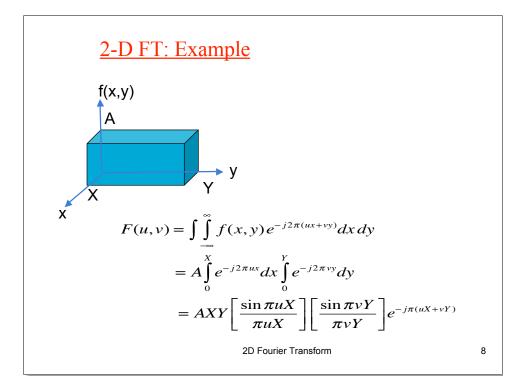


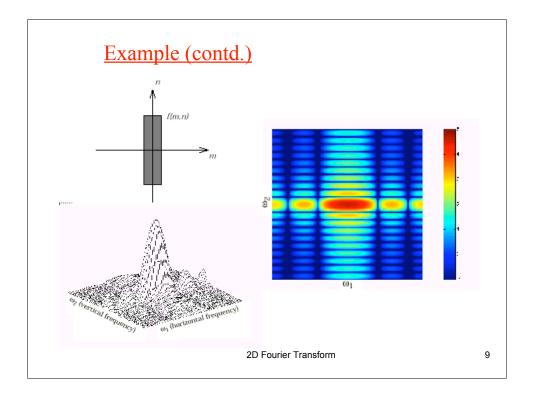


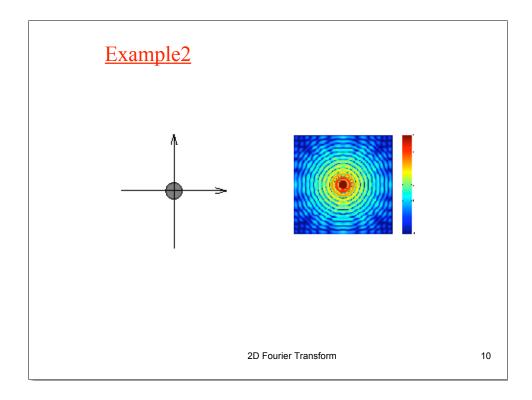


Consider a LSI system with impulse response h(x,y). Its output to the complex exponential is

$$g(x,y) = \iint_{-\infty}^{\infty} h(x-s,y-t)e^{j2\pi(us+vt)}ds dt$$
$$= \iint_{-\infty} h(\overline{x},\overline{y})e^{j2\pi(ux+vy)}e^{-j2\pi(u\overline{x}+v\overline{y})}d\overline{x} d\overline{y}$$
$$= H(u,v)e^{j2\pi(ux+vy)}$$
2D Fourier Transform







## **Discrete Fourier Transform**

Consider a sequence {u(n), n=0,1,2,...., N-1}. The DFT of u(n) is

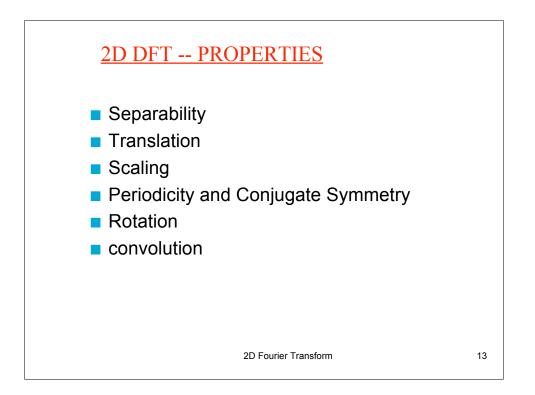
$$v(k) = \sum_{n=0}^{N-1} u(n) W_N^{kn}, \qquad k = 0, 1, \dots, N-1$$

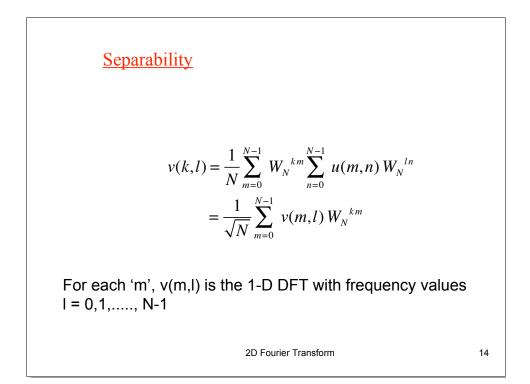
Where  $W_N = e^{-\sqrt{N}}$ , and the inverse is given by

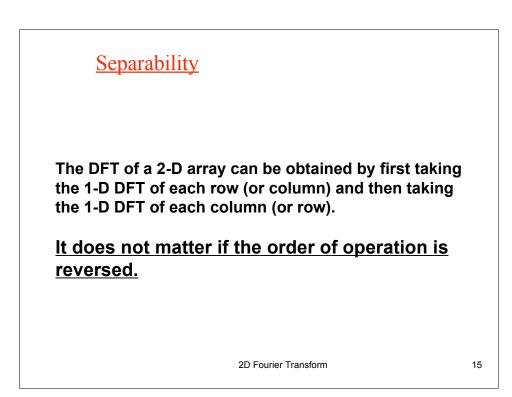
11

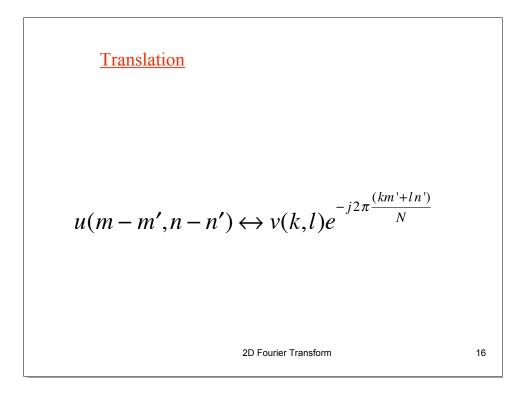
$$u(n) = \frac{1}{N} \sum_{k=0}^{N} v(k) W_N^{-kn}, \quad n = 0, 1, ..., N-1$$
2D Fourier Transform

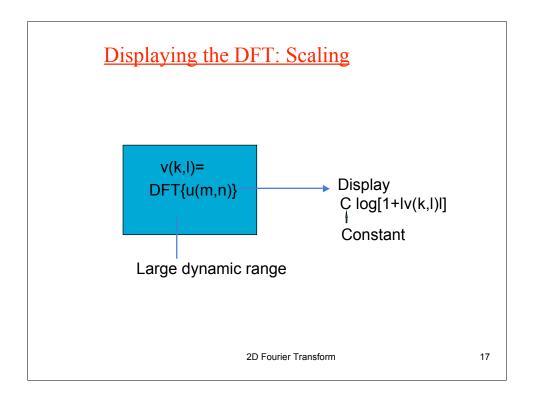
 $\begin{array}{l}<section-header> \textbf{DET}\\ \textbf{Differs it is convenient}\\ \textbf{to consider a}\\ \textbf{symmetric transform:} \end{array} \quad \begin{bmatrix} v(k) = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} u(n) W_N{}^{kn} & \text{and} \\ u(n) = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} v(k) W_N{}^{-kn} \\ u(n) = \frac{1}{\sqrt{N}} \sum_{n=0}^{N-1} \sum_{n=0}^{N-1} u(m,n) W_N{}^{km} W_N{}^{ln}, \\ u(m,n) = \frac{1}{N} \sum_{k=0}^{N-1} \sum_{l=0}^{N-1} v(k,l) W_N{}^{-km-ln} \\ u(m,n) = \frac{1}{N} \sum_{k=0}^{N-1} \sum_{l=0}^{$ 

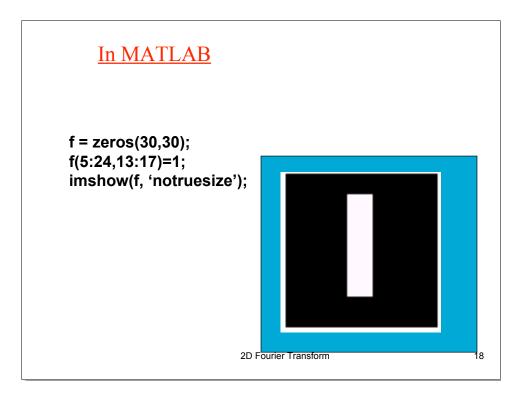


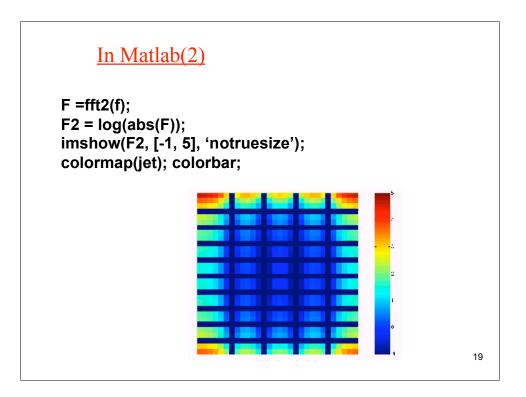


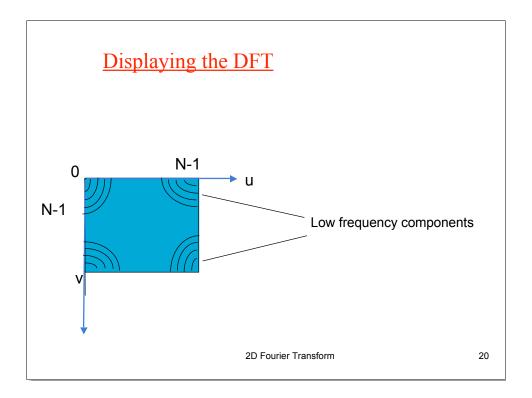










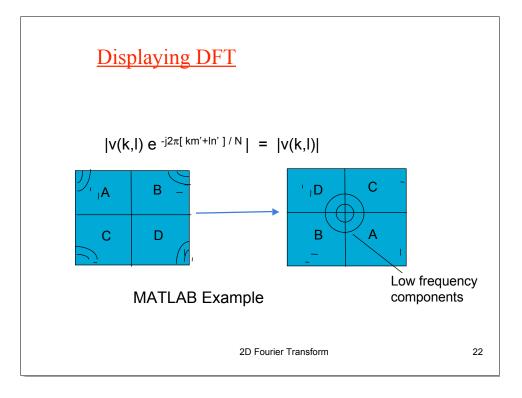


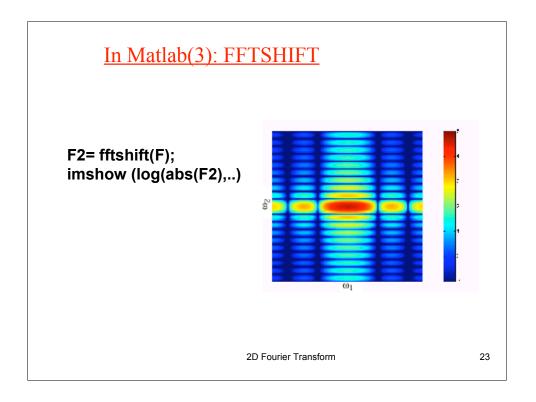


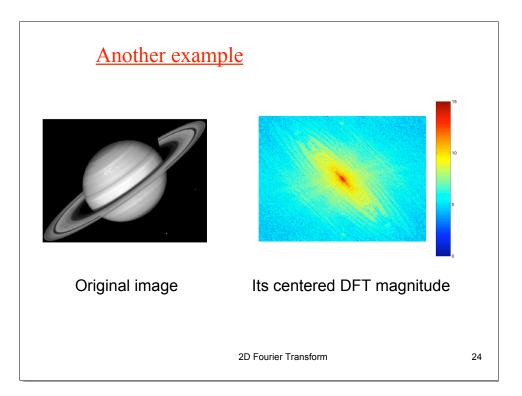
$$u(m,n)e^{\frac{j2\pi (k'm+l'n)}{N}} \leftrightarrow v(k-k',l-l') \text{ and}$$

$$u(m,n)(-1)^{m+n} \leftrightarrow v\left(k-\frac{N}{2}, l-\frac{N}{2}\right)$$
The origin of the F{u(m,n)} can be moved to the center of the array (N X N square) by first multiplying u(m,n) by (-1)^{m+n} and then taking the Fourier transform.  
Note: Shifting does not affect the magnitude of the Fourier transform.

2D Fourier Transform





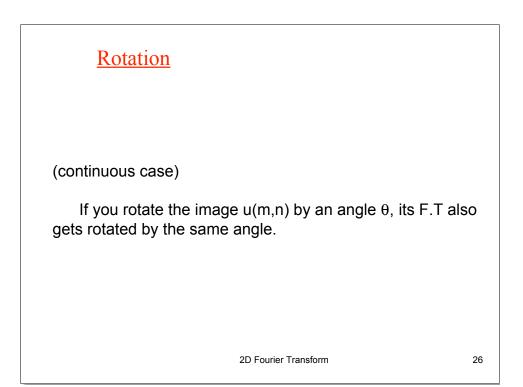


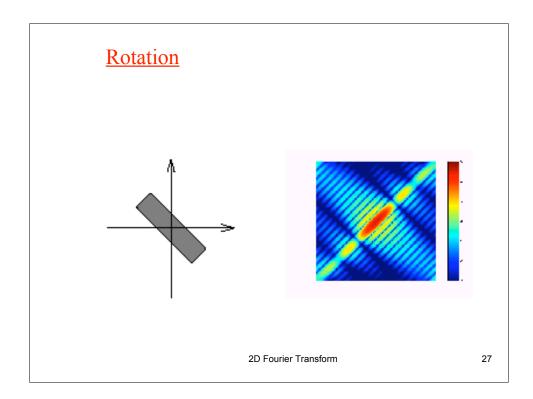
Periodicity & Conjugate Symmetry

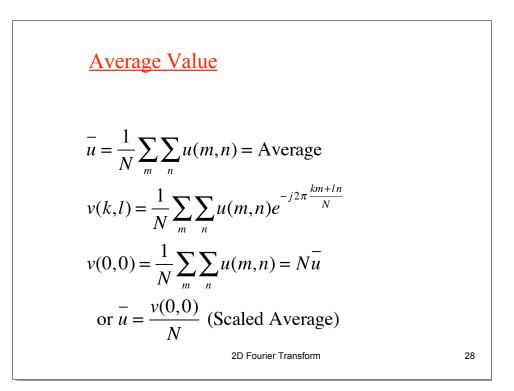
 $u(m,n) \xleftarrow{F} v(k,l)$ v(k,l) = v(k+N, l) = v(k, l+N) = v(k+N, l+N)

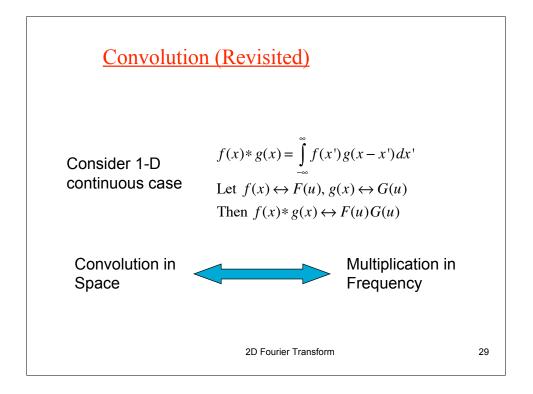
If u(m,n) is real, v(k,l) also exhibits conjugate symmetry v(k,l) = v\* (-k, -l) or |v(k,l)| = |v(-k, -l)|

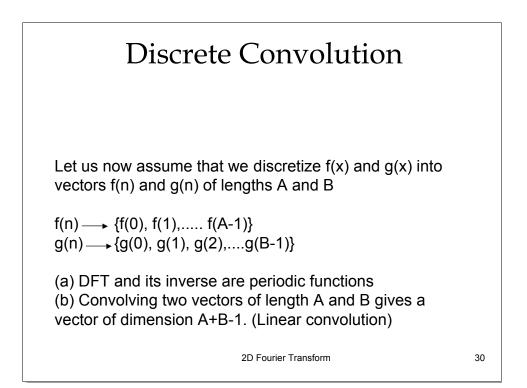
2D Fourier Transform

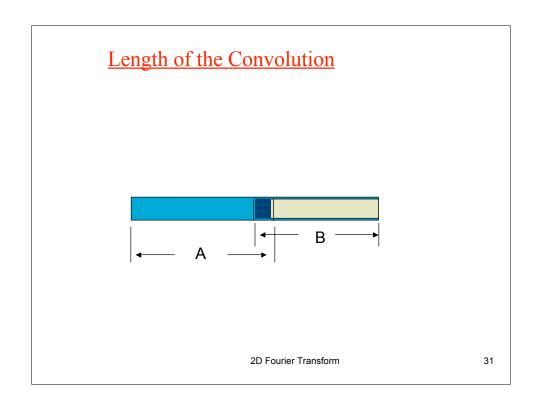


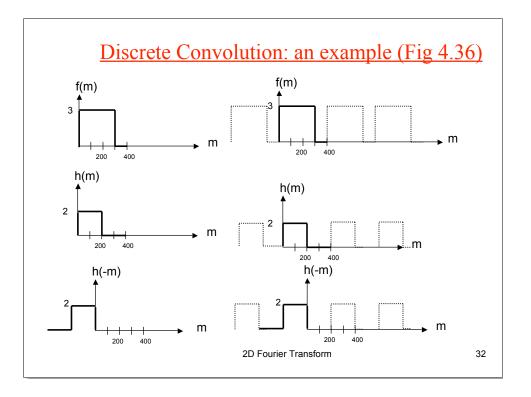


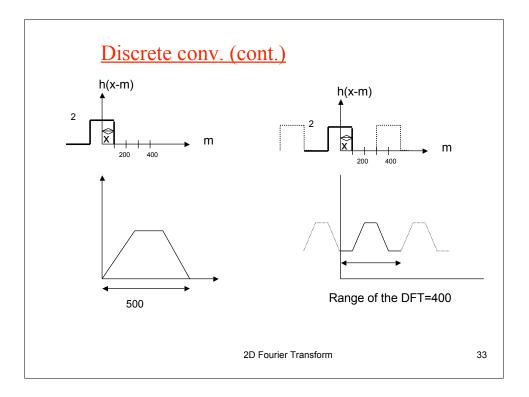


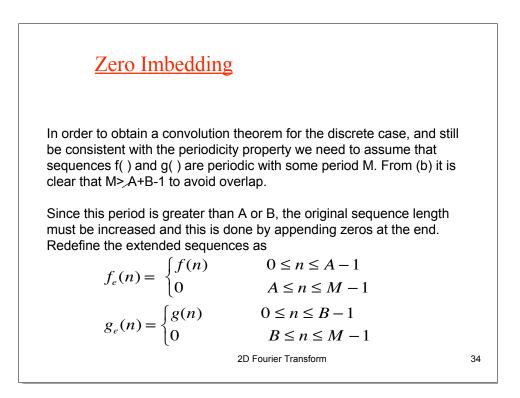


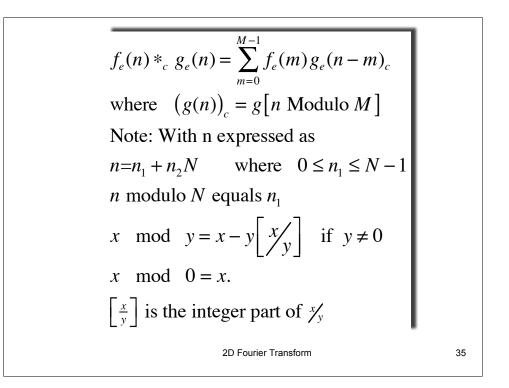






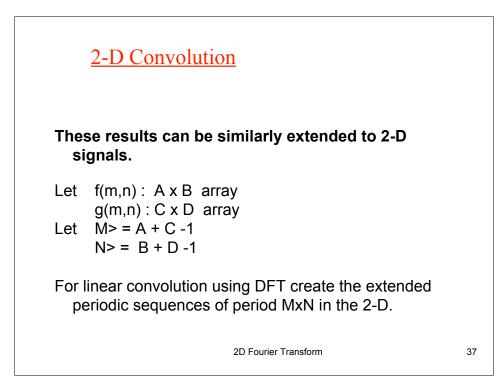


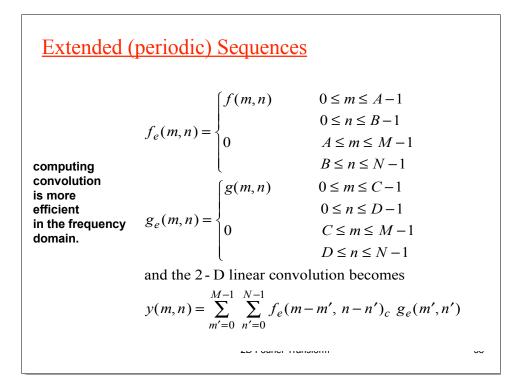


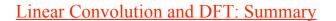


## TheoremThe DFT of the circular convolution of two sequences of<br/>length N is equal to the product of their DFTs.If $y(n) = \sum_{m=0}^{N-1} f(n-m)_c g(n)$ then<br/> $DFT[y(n)]_N = DFT[f(n)]_N DFT[g(n)]_N$ A linear convolution of two sequences can be obtained via<br/>FFT by embedding it into a circular convolution.

2D Fourier Transform







y(n) = f(n) \* g(n)

1. Let  $M \ge A+B-1$  be an integer for which the FFT algorithm is available.

2. Define the zero extended sequences  $f_e(n)$ ,  $g_e(n)$ .

3. Let  $F_e(k) = DFT \{ f_e(n) \}_M$ ,  $G_e(k) = DFT \{ g_e(n) \}_M$ . Let  $Y_e(k) = F_e(k)G_e(k)$ 

4. Take the I-DFT of  $Y_e(k)$  to obtain  $Y_e(n)$ . Then Y (n) =  $Y_e(n)$  for 0 <= n <= A+B-1

2D Fourier Transform

