

# Wireless Communications

## Lecture 2

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### [Trigonometry relationship]

$$\begin{aligned}\sin A \cos B &= \frac{\sin(A+B) + \sin(A-B)}{2} \\ \sin A \sin B &= -\frac{\cos(A+B) - \cos(A-B)}{2} \\ \cos A \cos B &= \frac{\cos(A+B) + \cos(A-B)}{2}\end{aligned}$$

$$\begin{aligned}\sin(A+B) &= \sin A \cos B + \cos A \sin B \\ \cos(A+B) &= \cos A \cos B - \sin A \sin B\end{aligned}$$

$$\begin{aligned}\sin(2A) &= 2 \sin A \cos A \\ \cos(2A) &= \cos^2 A - \sin^2 A = 1 - 2 \sin^2 A = 2 \cos^2 A - 1\end{aligned}$$

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### [Fourier Transform]

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi ft} dt$$

$$\begin{aligned}x(t) &\longrightarrow X(f) \\ x^*(t) &\longrightarrow X^*(-f)\end{aligned}$$

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### [Relating energy of $s(t)$ to its baseband equivalent $u(t)$ ]

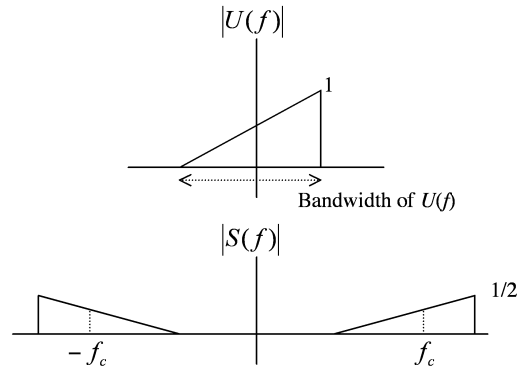
$|S(f)|$  symmetric around zero?

Finding  $S(f)$

$$\mathcal{F}\{u(t)e^{j2\pi f_c t}\} = U(f - f_c)$$

$$\begin{aligned}\operatorname{Re}\{x(t)\} &= \frac{x(t) + x^*(t)}{2} \\ \mathcal{F}\{u(t)e^{j2\pi f_c t}\} &= \frac{U(f - f_c) + U^*(-f - f_c)}{2}\end{aligned}$$

[example]



$$s(t) \longrightarrow \boxed{h(t)} \longrightarrow r(t) = h(t) * s(t) \longrightarrow R(f) = H(f)S(f)$$

$$\begin{aligned}S(f) &= \frac{1}{2} [U(f - f_c) + U^*(-f - f_c)] \\ H(f) &= [H_l(f - f_c) + H_l^*(-f - f_c)] \\ R(f) &= H(f)S(f) \\ &= \frac{1}{2} [U(f - f_c)H_l(f - f_c) + U^*(-f - f_c)H_l^*(-f - f_c)] \\ r(t) &= \operatorname{Re}\{z(t)e^{j2\pi f_c t}\}, \quad z(t) = \mathcal{F}^{-1}\{U(f)H_l(f)\}\end{aligned}$$

$$z(t) = u(t) * h_l(t) \Rightarrow r_l(t) = z(t)$$

$$u(t) \longrightarrow \boxed{h_l(t)} \longrightarrow r_l(t)$$

[Two ray model, proof of large  $d$ ]

$$r(t) = \text{Re} \left\{ u \left( t - \frac{l}{c} \right) e^{j2\pi f_c(t-l/c)} \right\} \frac{\lambda\sqrt{G_t G_r}}{4\pi l} + \text{Re} \left\{ u \left( t - \frac{x+x'}{c} \right) e^{j2\pi f_c \left( t - \frac{(x+x')}{c} \right)} \right\} \frac{\lambda\sqrt{G_t G_r}}{4\pi(x+x')} R$$

where  $R$  is the ground reflection coefficient.

Assuming wide enough time-duration for  $u(t)$ , such that  $u \left( t - \frac{l}{c} \right)$  and  $u \left( t - \frac{x+x'}{c} \right) \simeq u(t)$  and taking the time that the direct path arrived at origin:

$$r(t) = \text{Re} \left\{ u(t) e^{j2\pi f_c(t)} \right\} \frac{\lambda\sqrt{G_t G_r}}{4\pi l} + \text{Re} \left\{ u(t) e^{j2\pi f_c \left( t + \frac{l}{c} - \frac{(x+x')}{c} \right)} \right\} \frac{\lambda\sqrt{G_t G_r}}{4\pi(x+x')} R$$

The received power is

$$P_r = \left( \frac{\lambda}{4\pi} \right)^2 G_t G_r \left| \frac{u(t)}{l} + \frac{u(t)}{x+x'} e^{-j\Delta\phi} R \right|^2 \times \frac{1}{2}$$

Transmitted power  $P_t = P_u/2$

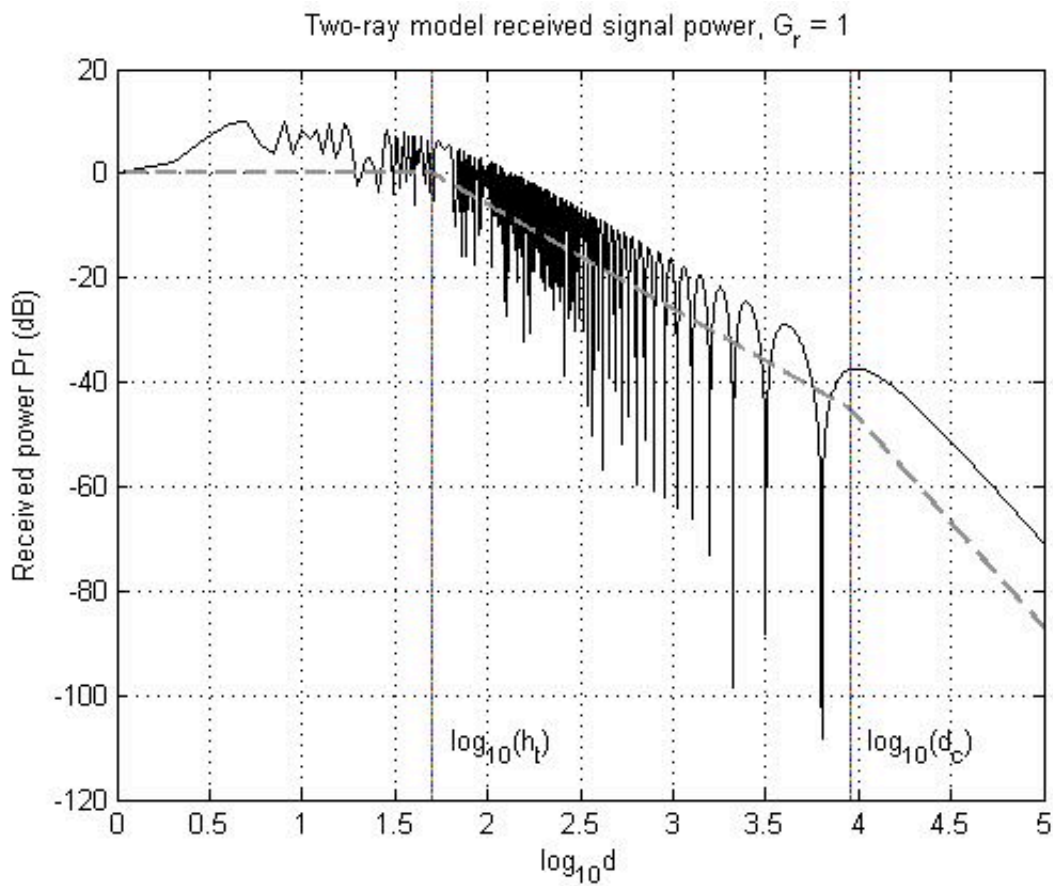
$$P_r = \left( \frac{\lambda}{4\pi} \right)^2 G_t G_r P_t \left| \frac{1}{l} + \frac{e^{-j\Delta\phi} R}{x+x'} \right|^2$$

If  $d$  denotes the horizontal separation of the antennas,  $h_t$  the transmitter height, and  $h_r$  the receiver height

$$\begin{aligned} \Delta\phi &= -2\pi f_c \left( \frac{l}{c} - \frac{(x+x')}{c} \right) \\ &= -2\pi \left( \frac{l}{\lambda} - \frac{(x+x')}{\lambda} \right), \quad \lambda = \frac{c}{f_c} \\ (x+x') - l &= \sqrt{(h_t + h_r)^2 + d^2} - \sqrt{(h_t - h_r)^2 + d^2} \\ d \gg h_t + h_r &\Rightarrow \Delta\phi \simeq \frac{4\pi h_t h_r}{\lambda d} \quad \text{using Taylor series approximation} \\ \sqrt{1+x^2} &\simeq 1 + \frac{x^2}{2} \\ R &= \frac{\sin\theta - Z}{\sin\theta + Z}, \quad \text{for large } d, \theta \simeq 0, R \simeq -1, x+x' \simeq l \simeq d, \Delta\phi \text{ is small} \end{aligned}$$

The the received power is

$$\begin{aligned}
 P_r &= \left(\frac{\lambda}{4\pi}\right)^2 G_t G_r P_t \left|\frac{1}{l}\right|^2 |1 - e^{-j\Delta\phi}|^2 \\
 &= \left(\frac{\lambda}{4\pi d}\right)^2 G_t G_r P_t |\Delta\phi|^2 \\
 &= G_t G_r \left[\frac{h_t h_r}{d^2}\right]^2 P_t
 \end{aligned}$$



$$f = 900 \text{ MHz}, R = -1, h_t = 50\text{m}, h_r = 2 \text{ m}, G_t = 1, \text{ and } G_r = 1$$