

Wireless Communications

Lecture 5

[Review of log-normal fading] $\psi_{\text{dB}} = 10 \log |\overline{ch}|^2$,

$$P_{\psi_{\text{dB}}} = \frac{1}{\sqrt{2\pi}\sigma_{\text{dB}}} e^{-\frac{(\psi_{\text{dB}} - m_{\text{dB}})^2}{2\sigma_{\text{dB}}^2}}, \quad m_{\text{dB}} = \alpha_{\text{dB}} - 10n \log d$$

[Example] Best path loss fit is $-31.54 \text{ dB} - 37.1 \log d$ for the data below. Find σ_{dB}^2 , variance of log-normal fading

d	P_r/P_t
10 m	-70 dB
20 m	-75 dB
50 m	-90 dB
100 m	-110 dB
300 m	-125 dB

[Solution] Take the average out and average the rest for variance:

$$\begin{aligned} \sigma_{\text{dB}}^2 &= \frac{1}{5} [(-70 + 31.54 + 37.1 \log 10)^2 + \\ &\quad + (-75 + 31.54 + 37.1 \log 20)^2 + \dots] \\ &= 13.29 \\ \sigma_{\text{dB}} &= 3.65 \end{aligned}$$

[Outage Probability] $\text{prob}\{P_{r_{\text{dB}}} \leq P_{\min}\}$

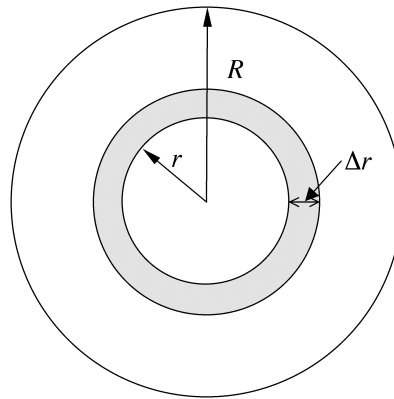
$$\begin{aligned} P_{r_{\text{dB}}} &= P_{T_{\text{dB}}} + \psi_{\text{dB}} \quad (\text{Normal distribution}) \\ \text{prob}\{P_{r_{\text{dB}}} \leq P_{\min}\} &= 1 - Q\left(\frac{P_{\min} - \overbrace{(P_{T_{\text{dB}}} + K_{\text{dB}} - 10\gamma \log d)}^{\alpha_{\text{dB}} - 10n \log d}}{\sigma_{\psi_{\text{dB}}}}\right) \\ Q(z) &= \frac{1}{\sqrt{2\pi}} \int_z^\infty e^{-x^2/2} dx \end{aligned}$$

[Example 2.5 of Goldsmith's] Find outage probability at 150 m, $P_{\min} = -110.5$ dBm, $P_t = 10$ mW for previous example.

[solution] We have $K = -31.54$, $\gamma = 3.71$. Therefore,

$$P_{\text{outage}} = 1 - Q\left(\frac{-110.5 - (10 - 31.54 - 37.11 \log(150))}{3.65}\right) = 0.0121$$

[Cell Coverage] Find what percentage of the cell is above certain level?



We know $\text{prob}\{P_r(r) \geq P_{\min}\}$, then:

$$\begin{aligned} C_{r,r+\Delta r} &= \text{area in the ring above } P_{\min} \\ &= \text{prob}\{P_r(r) \geq P_{\min}\} [\pi(r + \Delta r)^2 - \pi r^2] \\ &= \text{prob}\{P_r(r) \geq P_{\min}\} 2\pi r \Delta r \\ C_{0,R} &= 2\pi \int_0^R r \underbrace{\text{prob}\{P_r(r) \geq P_{\min}\}}_{Q\left(\frac{P_{\min} - (P_{T_{\text{dB}}} + K_{\text{dB}} - 10\gamma \log r)}{\sigma_{\text{dB}}}\right)} dr \\ C &= \frac{C_{0,R}}{\pi R^2} = \text{percentage of area above } P_{\min} \end{aligned}$$

Calculation involves $C = \frac{2}{R^2} \int_0^R rQ(a + b \ln r)dr$, $dr/r = du \rightarrow \ln r = u, \ln 0 = -\infty$ and integration by parts

$$\begin{aligned}
 \int_{-\infty}^{\ln R} e^{2u} Q(a + bu) du &= \frac{1}{2} e^{2u} Q(a + bu) \Big|_{-\infty}^{\ln R} + \int_{-\infty}^{\ln R} \frac{(1/2 e^{2u}) b}{\sqrt{2\pi}} e^{-\frac{(a+bu)^2}{2}} du \\
 &= \frac{1}{2} e^{2u} Q(a + bu) \Big|_{-\infty}^{\ln R} + \frac{b}{2\sqrt{2\pi}} \int_{-\infty}^{\ln R} e^{2u - \frac{(a+bu)^2}{2}} du \\
 &= \dots \\
 &= Q(a) + e^{\frac{2-2ab}{b^2}} Q\left(\frac{2-ab}{b}\right)
 \end{aligned}$$

[Rician channel] $ch(t) = ch_I(t) + jch_Q(t)$

We assume uniform angle of arrival $\Rightarrow ch_I(t)$ and $ch_Q(t)$ are zero mean. If a LOS path exists: separate that part

$$ch(t) = \underbrace{z_0(t)}_{\text{LOS}} + \underbrace{z(t)}_{\text{zero mean}}$$

$r(t) = |ch(t)|$ has a Rician distribution

$$P(r) = \frac{r}{\sigma^2} e^{-\frac{(r^2+s^2)}{2\sigma^2}} I_0\left(\frac{rs}{\sigma^2}\right), \quad r \geq 0, I_0(x) : \text{modified Bessel function}$$

where $2\sigma^2 = \sum_{n \neq \text{LOS path}} \overline{\alpha_n^2}$ is the average power in the non-LOS multipath components and $s^2 = |z_0|^2$ is the power in the LOS component.

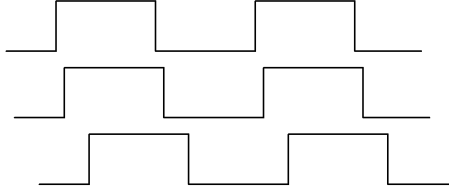
$$\mathbf{E}[r^2(t)] = s^2 + 2\sigma^2$$

The fading parameter K is defined by

$$K = \frac{s^2}{2\sigma^2}$$

K is a measure of the severity of the fading.

Resolvable Paths:



[Wideband Channels] The received signal is

$$r(t) = \text{Re} \left\{ \sum_{n=1}^{N(t)} \alpha_n(t) u(t - \tau_n(t)) e^{j2\pi(f_c + \Delta f_n(t))t - j2\pi f_c \tau_n(t)} \right\}$$

- The different multipath components can be resolved.
- This results in *Intersymbol Interference* (ISI)
- The approximation $u(t - \tau_n(t)) \approx u(t)$ is no longer valid.

$$r(t) = \text{Re} \left\{ \sum_{n=1}^{N(t)} \alpha_n(t) e^{j\phi_n(t)} u(t - \tau_n(t)) e^{j2\pi f_c t} \right\}$$

$$r_{bb}(t) = \sum_{n=1}^{N(t)} \alpha_n(t) e^{j\phi_n(t)} u(t - \tau_n(t))$$

$$u(t) \longrightarrow \boxed{\sum_{n=1}^{N(t)} \alpha_n(t) e^{j\phi_n(t)} \delta(t - \tau_n(t))} \longrightarrow r_{bb}(t)$$

[Note]: Two time progression

1. distance R approach.

$$h(t, R) = \sum \alpha_n(R) e^{j\phi_n(R)} \delta(t - \tau_n(R))$$

2. t, τ approach

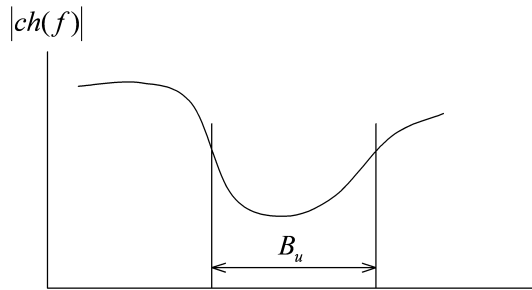
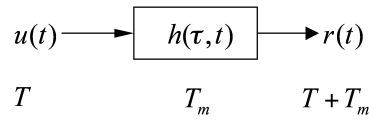
$$h(\tau, t) = \sum \alpha_n(t) e^{j\phi_n(t)} \delta(\tau - \tau_n(t))$$

[Example 3.1 of Goldsmith's] Consider a TX in a factory near a conveyor belt. LOS path with α_0, ϕ_0 . Every T_0 second a metal item comes down the belt with α_1, ϕ_1, τ_1 . τ_1 is delay difference. Characterize the channel.

$$h(\tau, t) = \begin{cases} \alpha_0 e^{j\phi_0} \delta(\tau) & t \neq nT_0 \\ \alpha_0 e^{j\phi_0} \delta(\tau) + \alpha_1 e^{j\phi_1} \delta(\tau - \tau_1) & t = nT_0 \end{cases}$$

[Increase in time length]

$T_m \gg T \Rightarrow$ resolvable. Equivalently, $B_u \ll T_m^{-1}$ which is called is wideband channel.



\Rightarrow wideband channel or frequency selective fading.

[Example]

