LINEAR FIR ADAPTIVE FILTERING (II) Stochastic Gradient-based Algorithms: (Least-Mean-Square [LMS])

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• Why the LMS Adaptive Filter ?

• Steepest descent algorithm has been used to obtain an iterative solution to fixed normal equations.

• We need to design a filter which is responsive to changes in the input signal environment, that is we need an interactive structure that is dependent on the input data.

• It is possible to construct a steepest-descent algorithm by replacing the fixed auto and cross-correlation matrices by their time dependent equivalents.

 $\underline{w}(n+1) = w(n) + \mu[\underline{p}(n) - \underline{R}(n)w(n)]$

We note two main problems with this approach:

A. It is very computationally intensive and expensive since we have to compute $\underset{\sim}{\mathbb{R}}$ and $\underset{\sim}{\mathbb{p}}$ at each point (n).

B. It may not be possible to compute $\underset{\sim}{R(n)}$ and $\underset{\sim}{p(n)}$ if only a single realization of the process is available.

LMS algorithm is based on estimating the gradient of the mean-squared error by the gradient of the instantaneous value of the squared error.

$$\nabla(J(n)) = \frac{\partial J}{\partial \underline{w}(n)} = \frac{\partial E\{e^2(n)\}}{\partial \underline{w}(n)}$$

Is replaced by:
$$\hat{\nabla}(J(n)) = \frac{\partial \hat{J}}{\partial \underline{w}(n)} = \frac{\partial e^2(n)}{\partial \underline{w}(n)}$$





(b) FIGURE 5.1 (continued on the next page)



FIGURE 5.1 (a) Block diagram of adaptive transversal filter. (b) Detailed structure of the transversal filter component. (c) Detailed structure of the adaptive weight-control mechanism.

Notes on Previous Figure:

Fig 5.1a Adaptive transversal filter consists of a transversal filter around which the LMS algorithm is built, note also a mechanism for performing the adaptive control process on the tap weights of the transversal filter.

Fig 5.1b The tap input vector is:

 $\underline{u}(n)$: M-1 is the number of delay elements.

 $\underline{\mathbf{w}}(\mathbf{n})$: Mx1 tap weigth vectors

Fig 5.1c Note the correction $\delta \hat{w}_k(n)$ applied to the tap weight $\hat{w}_k(n)$ at time n+1. μ is called the adaptation constant or step-size parameter. The Least-Mean-Square (LMS) algorithm

From the steepest-descent algorithm, reproduce:

$$\nabla(J(n)) = -2\underline{p} + 2\underline{R}w(n)....(1)$$
$$\underline{w}(n+1) = \underline{w}(n) + \frac{1}{2}\mu[-\nabla J(n)]....(2)$$

Define:

$$\underline{R}(n) = \underline{u}(n)u^{H}(n)....(3)$$
$$\hat{p}(n) = \underline{u}(n)d^{*}(n)....(4)$$

By using (3) and (4) in (1):

 $\hat{\nabla}(J(n)) = -2\underline{u}(n)d^*(n) + 2\underline{u}(n)\underline{u}^H(n)\overline{W}(n)...(5)$

The instantaneous estimate of the gradient vector.

Putting equation (5) in quation (2):

 $\underline{\hat{w}}(n+1) = \underline{\hat{w}}(n) + \mu \underline{u}(n) [d^*(n) - u^H(n) \underline{\hat{w}}(n)]...(6)$

LMS Algorithm: Three Basic Steps

1. Filter Output:

$$y(n) = \hat{w}^{H}(n)\underline{u}(n)....(7a)$$

2. Estimation of errors:

$$e(n) = d(n) - y(n)....(7b)$$

3. Tap weight adaptation: $\hat{w}(n+1) = \hat{w}(n) + \underbrace{\mu u(n)e^{*}(n)...(7c)}_{\text{Correction Term}}$ The algorithm gets initiated with: $\hat{w}(0) = 0$



Notes on the LMS algorithms from the above figure: $\underline{\hat{w}}(n)$

1.) Algorithm is multivariable since has a dimension greater then 1.

2.) Nonlinear since outer feedback loop depends on tap^n input vector.

3.) becomes a random vector fro n > 0.

Stability Analysis of the LMS algorithm:

As a rule of thumb, we have to choose μ so that the following two forms of convergence are satisfied:

Convergence in the mean:

 $E[\hat{w}(n)] \rightarrow \underline{W}_0 \text{as } n \rightarrow \infty$

Convergence in the mean square:

 $J(n) \rightarrow J(\infty)$ as $n \rightarrow \infty$

where $J(\infty) > J_{\min}$

and $J(\infty)$ is finite

 J_{min} corresponds to the weiner solution.

Fundamental assumption for analysis of the LMS algorithm.

1. The tap input vectors $\underline{u}(1)$, $\underline{u}(2) \dots \underline{u}(n)$ constitute a sequence of statistically independent vectors.

2. At time n, the tap-input vector $\underline{u}(n)$ is statistically independent of all previous samples of the desired response d(1), d(2),....d(n-1).

3. At time n, the desired response d(n) is dependent on the $\underline{u}(n)$, but statistically independent of all previous samples of the desired response.

4. u(n) and d(n) consist of mutually gaussiandistributed random variables for all n. With the assumption of mutually gaussian distributed random variables in (4), (1) and (2) are equivalent to conditions of uncorrelatedness:

 $E[\underline{u}(n)u^{H}(k)] = \underline{0}....k = 0,1,...n-1$ $E[\underline{u}(n)d^{*}(k)] = 0....k = 0,1,...n-1$

the assumption of independence theory [(1) -(3)] is more general when (4) is not assumed.

Average tap weight analysis

Define: $\underline{\varepsilon}(n) = \hat{w}(n) - \underline{w}(0)$ the weight error vector

where $\underline{\hat{w}}(n)$ the estimate produced by the LMS algorithm \underline{W}_0 the optimum weiner solution.

Start from:

$$\hat{\underline{w}}(n+1) = \hat{\underline{w}}(n) + \mu \underline{u}(n)[d^*(n) - u^H(n)\hat{\underline{w}}(n)]$$
Rewrite:

$$\hat{\underline{w}}(n+1) - \underline{w}_0 = \hat{\underline{w}}(n) - \underline{w}_0 + \mu \underline{u}(n)[d^*(n) - u^H(n)\hat{\underline{w}}(n)]$$

$$\underbrace{\hat{\underline{v}}(n+1) = \hat{\underline{v}}(n) + \mu \underline{u}(n)[d^*(n) - u^H(n)\hat{\underline{w}}(n)]$$

$$\underline{\underline{\varepsilon}}(n+1) = \underline{\varepsilon}(n) + \mu \underline{u}(n)[d^*(n) - u^H(n)\hat{\underline{w}}(n)]$$

$$\underbrace{\underline{\varepsilon}(n+1) = [I - \mu \underline{u}(n)u^H(n)]\underline{\varepsilon}(n) + \mu \underline{u}(n)e_0^*(n)$$

$$\underline{e}(n) = d(n) - \hat{w}^H(n)\underline{u}(n)$$

$$= d(n) - w_0^H \underline{u}(n) - \underline{\varepsilon}^H(n)\underline{u}(n)$$

$$= e_0(n) - \underline{\varepsilon}^H(n)\underline{u}(n)$$

 e_0 : the estimation error produced in the optimum weiner solution.

Now by taking the mathematical expectation: $E[\underline{\varepsilon}(n+1)] = E[[I - \mu \underline{u}(n)u^{H}(n)]\underline{\varepsilon}(n)] + \mu [E\underline{u}(n)e_{0}^{*}(n)]$

Since $\underline{\varepsilon}(n)$ and $\underline{u}(n)$ are independent, the first term:

$$E[[I - \mu \underline{u}(n)u^{H}(n)]\underline{\varepsilon}(n)]$$

= $[I - \mu E[\underline{u}(n)u^{H}(n)]]E[\underline{\varepsilon}(n)]$
= $[I - \mu \mathbf{R}]E[\underline{\varepsilon}(n)]$

Where:

$$\underbrace{R} = E[\underline{u}(n)\underline{u}^{H}(n)]$$

From the orthogonality principle:

$$E[\underline{u}(n)e_0^*(n)] = \underline{0}$$

 $E[\underline{\varepsilon}(n+1)] = [I - \mu \underline{R}]E[\underline{\varepsilon}(n)].....(A)$

Compare equation (A) with that we obtained in the case of steepest-decent algorithm where:

$$\underline{c}(n+1) = [I - \mu \underline{R}]\underline{c}(n)$$

We therefore conclude from (A) that:

the mean of $\underline{\varepsilon}(n)$ converges to zero as $n \to \infty$, provided:

$$0 < \mu < \frac{2}{\lambda_{max}}$$
 (B)

Conclusion: if μ is set as in B

 $E[\underline{\hat{w}}(n)] \to W_0 \qquad \text{as } n \to \infty$

that is: the LMS algorithm is convergent in the mean.

Mean-Squared-error Analysis:

Start From:

$$e(n) = d(n) - \underline{\hat{w}}^{H}(n)\underline{u}(n)$$

= $d(n) - \underline{w}^{H}_{0}\underline{u}(n) - \underline{\varepsilon}^{H}(n)\underline{u}(n)$
= $e_{0}(n) - \underline{\varepsilon}^{H}(n)\underline{u}(n)$

Now:

$$J(n) = E[|e(n)|^{2}]$$

= $E[(e_{0}(n) - \underline{\varepsilon}^{H}(n)\underline{u}(n))(e_{0}^{*}(n) - \underline{\varepsilon}(n)\underline{u}^{H}(n))]$
= $J_{\min} - E[\varepsilon^{H}(n)\underline{u}(n)\underline{\varepsilon}(n)\underline{u}^{H}(n)]$

Where J_{min} is the minimum mean-squared error produced by the optimum weiner filter.

The second term can be manipulated to:

 $E[\varepsilon^{H}(n)\underline{u}(n)\underline{\varepsilon}(n)\underline{u}^{H}(n)]$ = $tr\{E[\varepsilon^{H}(n)\underline{u}(n)\underline{\varepsilon}(n)\underline{u}^{H}(n)]\}$ = $tr\{\underbrace{R}\underset{\sim}{\mathcal{K}}(n)\}$

$$J(n) = J_{\min} - tr[\underline{RK}(n)] \quad (C)$$

Where \underline{R} : correlation matrix
 \underline{K} : the weight-error correlation matrix

Conclusion: since $tr\{\mathcal{R},\mathcal{K}(n)\}$ is positive definite for all *n*. LMS allways produces a mean-squared error J(n) that is in excess of the minimum-mean-squared-error J_{min}

Rewrite equation (C):

$$J(n) = J_{\min} - tr\{\mathcal{R}_{\mathcal{K}}(n)\}$$
(C)

= the excess mean squared error.

By transforming to eigen-coordinate system: $\underline{Q}^{H} \overset{R}{\approx} \overset{Q}{\approx} = \overset{\Lambda}{\approx}$

and defining:

$$Q^{H} \underbrace{K}(n) \underbrace{Q} = \underbrace{X}(n)$$

We can write:

 $tr[\underbrace{R}_{ex} \underbrace{K}(n)] = tr[\underbrace{\Lambda}_{ex} \underbrace{X}(n)]$ $J_{ex}(n) = tr[\underbrace{\Lambda}_{ex} \underbrace{X}(n)] \quad (D)$

In diagonal coordinate system:

$$J_{ex}(n) = \sum_{i=1}^{M} \lambda_i x_i(n) \qquad (E)$$

where $x_i(n)$, I=1,2,...,M, diagonal elements of $\underline{X}(n)$.

Equation (E) can be re-written as:

$$J_{ex}(n) = \underline{\lambda}^T x(n)$$

Where
$$\underline{x}(n)$$
 satisfies
 $\underline{x}(n+1) = \underline{B} \underline{x}(n) + \mu^2 J_{\min} \underline{\lambda}$ (F)
where \underline{B} is the MxM matrix with:
 $b_{ij} = \begin{cases} (1-\mu\lambda_i)^2 + \mu^2\lambda_i^2 \dots i = j \\ \mu^2\lambda_i\lambda_j \dots i \neq j \end{cases}$

The matrix B is real, positive and symmetric.

By solving the first order vector difference equation (F) gives:

$$\underline{x}(n) = \underline{\mathcal{B}}^{n} \underline{x}(0) + \mu^{2} J_{\min} \sum_{i=0}^{n-1} \underline{\mathcal{B}}^{i} \underline{\lambda} \quad (G)$$
$$\sum_{i=0}^{n-1} \underline{\mathcal{B}}^{i} = (\underline{I} - \underline{\mathcal{B}})^{n} (\underline{I} - \underline{\mathcal{B}})^{-1}$$

Then equation (G) becomes:

 $\underline{x}(n) = \underline{B}^{n} [\underline{x}(0) - \mu^{2} J_{\min} (\underline{L} - \underline{B})^{-1} \underline{\lambda}] + \mu^{2} J_{\min} (\underline{L} - \underline{B})^{-1} \underline{\lambda}$

transient Component

Steady state Component Since \underline{B} is symmetric, we can apply an orthogonal similarity transformation.

$$\underline{G}^T \underline{B} \underline{G} = \underline{C}$$

G: orthogonal matrix

$$\mathcal{G}^T \mathcal{G} = \mathcal{I}$$

$$\underline{B}^{n} = \underline{G} \underbrace{C}^{n} \underbrace{G}^{T} = \underline{C}$$

 \underline{C} : Diagonal matrix with C_i , $i = 1, 2, \dots, M$

 g_i : eigen vectors of B associated with eigenvalues c_i

We can then manipulate the equation for $\underline{x(n)}$ to:

$$\underline{x}(n) = \sum_{i=1}^{M} c_{i}^{n} \underline{g}_{i} \underline{g}_{i}^{T} [\underline{x}(0) - \underline{x}(\infty)] + \underline{x}(\infty)$$
(H)

Where for stability:

 $0 < c_i < 1$ for all i

$$x(\infty) = \mu^2 J_{\min} (\underline{L} - \underline{B})^{-1} \underline{\lambda}$$

And:

$$\underbrace{G} \underbrace{C}^{n} \underbrace{G}^{T} = \sum_{i=1}^{M} c_{i}^{n} \underbrace{g}_{i} \underbrace{g}_{i}^{T}$$

Now the excess mean-squared error:

$$J_{ex}(n) = \underline{\lambda}^{T} \underline{x}(n)$$

$$= \sum_{i=1}^{M} c_{i}^{n} \underline{\lambda}^{T} \underline{g}_{i} \underline{g}_{i}^{T} [\underline{x}(0) - \underline{x}(\infty)] + J_{ex}(\infty) \quad (I)$$
Where $J_{ex}(\infty) = \underline{\lambda}^{T} \underline{x}(\infty)$

$$= \sum_{i=1}^{M} \lambda_{j} x_{j}(\infty)$$

After some more algebra:

We may write an expression for:

$$J_{ex}(\infty) = J_{\min} \frac{\sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i}}{1 - \sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i}} \qquad (J)$$

Now Since:

$$J_{ex}(n) = J(n) - J_{min}$$

$$= tr\{\underline{R} \underbrace{K}(n)\}$$

Putting equations (J) in (I), the time evolution of the mean squared error for the LMS algorithm:

$$J(n) = \sum_{i=1}^{M} r_i C_i^n + \frac{Jim}{1 - \sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i}}$$
(K)

Where: $r_i = \underline{\lambda}^T \underline{g}_i g_i^T [\underline{x}(0) - \underline{x}(\infty)]$ $x_i(0) = E[q_i^H \underline{\varepsilon}(0) \varepsilon^H(0) \underline{q}_i] \dots i = 1, 2, \dots M$ $\underline{\varepsilon}(0) = \hat{w}(0) - \underline{w}(0)$ q: eigen vector of R Conclusions from equation (K)

a.) The first term is (K)

$$\sum_{i=1}^{M} \mathcal{V}_i C_i^n$$

Represents the transient component of J(n)where r_i are constants and c_i are the eigenvalues of <u>B</u> and they are real positive numbers since <u>B</u> is a real symmetric positive definite matrix.

b.) $J(n) \rightarrow J(\infty)$ if, and only if, μ : satisfies two conditions:

$$0 < \mu < \frac{2}{\lambda_{\max}}$$

$$\sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i} < 1$$

$$\lambda_i I=1,2,...,M, \text{ eigenvalues of } \underline{R} .$$
The LMS algorithm is convergent in the mean

square.

c. The mean squared error produced by the LMS algorithm has the final value:

$$J(\infty) = \frac{J_{\min}}{1 - \sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i}}$$

d. The misadjustment is defined as:

$$M = \frac{J_{ex}(\infty)}{J_{\min}}$$
$$M = \frac{\sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i}}{1 - \sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i}}$$

(i) If μ is small compared to $2/\lambda_{max}$

$$M \approx \frac{\mu}{2} \sum_{i=1}^{M} \lambda_i$$

(ii) By defining:

$$\lambda_{av} = \frac{1}{M} \sum_{i=1}^{M} \lambda_i$$

We can define the convergence time constant for the LMS algorithm:

$$(\tau)_{mse,av} \approx \frac{1}{2\mu\lambda_{av}}$$
$$M = \frac{\mu M \lambda_{av}}{2} \approx \frac{M}{4\tau_{mse,av}}$$

Note that we have conflicting requirements in that if μ is reduced to reduce M then $(\tau)_{mse av}$ is increased. Conversely if μ is increased to reduce $(\tau)_{mse av}$ then M is increased. The choice of μ becomes an important compromise.

For real data, it should be noted the real LMS algorithm in the mean square is give by:

$$\sum_{i=1}^{M} \frac{\mu \lambda_i}{2 - \mu \lambda_i} < 1$$