

# **Finite Impulse Response (FIR)**

## **Digital Filters (IV)**

### **Impulse Response Coefficients calculation with the Optimal method**

**Yogananda Isukapalli**

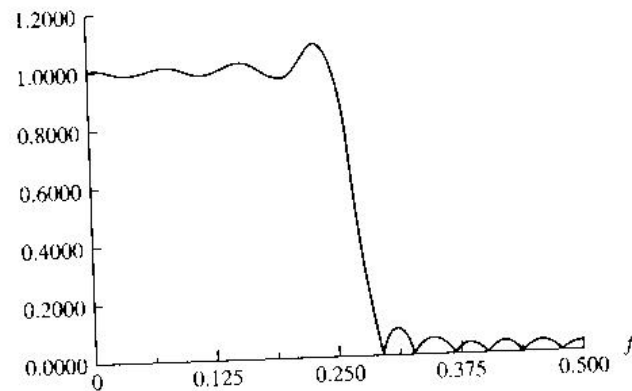
# Optimum Design of Filters

- Design of FIR filters by windowing is straightforward. However, we wish to design a filter that is the “best” that can be achieved for a given value of ‘ $M$ ’
- We know that the rectangular window provides the best mean-square approximation to the desired-frequency for a given value of  $M$ . However this approximation shows adverse behavior at discontinuities of  $H_d(e^{j\omega})$ .
- The window method although simple does not provide for individual control over approximation errors in different bands.

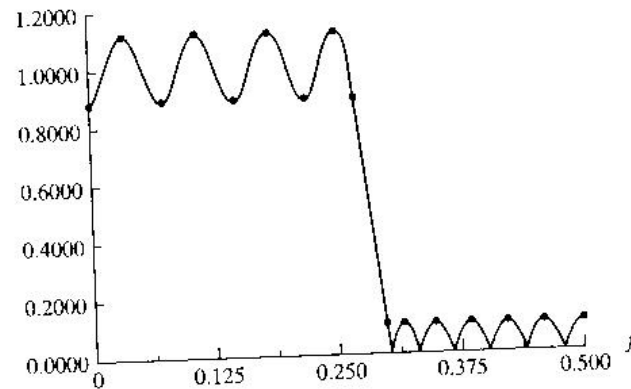
- Frequency selective filters designed by windowing often have the property that the *error is greatest on either side of a discontinuity* of the ideal frequency response, and the error becomes smaller for frequencies away from the discontinuity as shown in Fig 1a.
- But, if the ripples are distributed more evenly over the pass band and stop band, for example as shown in Fig 1b, a better approximation of the desired frequency response can be achieved.
- *The optimal method is hence based on the concept of equiripple pass band and stop band.*

# Optimum Design of Filters

contd...



(a)



(b)

Fig 1<sup>1</sup>. (a) In the window filter, the ripples are largest near the band edge  
(b) In the optimal filter, the ripples have the same peaks (equiripple) in the pass band or stop band.

- Consider the low pass filter frequency response shown in Fig 2a.
- In the pass band , the practical response oscillates between  $1 - \delta_p$  and  $1 + \delta_p$ .
- In the stop band, the filter response lies between 0 and  $\delta_s$ .
- The difference between the ideal filter and the practical response can be viewed as an error function (Fig 2b):

$$E(\omega) = W(\omega)[H_D(\omega) - H(\omega)]$$

where  $H_D(\omega)$  is the ideal or desired response and  $W(\omega)$  is a weighting function that allows the relative error of approximation between different bands to be defined.

# Optimum Design of Filters

contd...

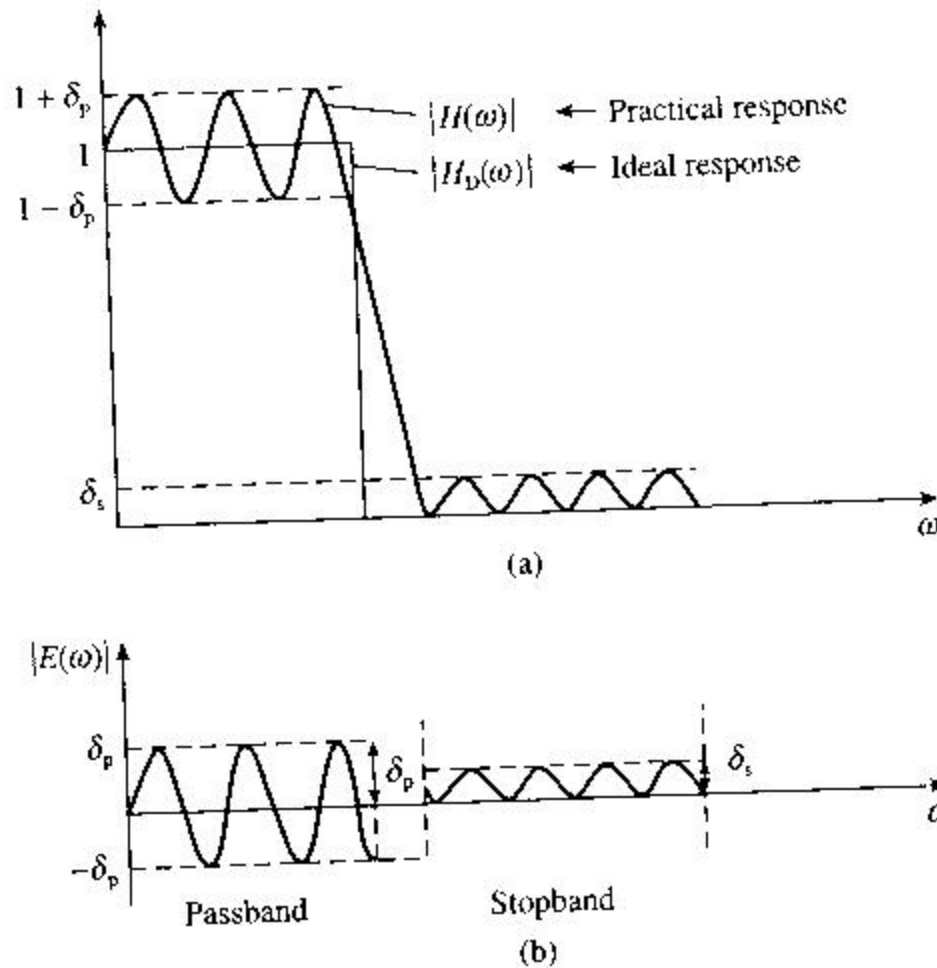


Fig 2<sup>1</sup>. (a) Frequency response of an optimal low pass filter (b) Response of the error between the ideal and practical responses ( $\delta_p = 2\delta_s$ )

- In the optimal method, the objective is to determine the filter coefficients,  $h(n)$ , such that the value of the maximum weighted error,  $|E(\omega)|$  is minimized in the pass band and stop band, i.e.

$$\min[\max|E(\omega)|] \quad \text{over the pass band and stop band.}$$

- Algorithmic techniques in designing filters to meet the above mentioned specifications must systematically vary the impulse response  $h[n]$  to obtain the best results.
- Design algorithms have been developed in which some parameters ( $M$ ,  $w_s$ ,  $w_p$ ,  $\delta_s$  and  $\delta_p$ ) are fixed and an iterative procedure is used to obtain the optimum adjustments of others.

## Parks-McClellan algorithm

- The most dominant of the optimum solution generation algorithms is the Parks-McClellan algorithm.
- The Parks-McClellan algorithm is based on reformulating the filter design problem as a problem in polynomial approximation.

Consider the design of a type I FIR filter having zero phase. Its frequency response is given by:

$$H(e^{j\omega}) = \sum_{n=0}^{N-1} h[n]e^{j\omega n}$$

$$\text{for } M = N / 2,$$

$$A(e^{j\omega}) = \sum_{n=-M}^M h[n]e^{j\omega n}$$

$$\text{i.e. } A(e^{j\omega}) = h[0] + \sum_{n=1}^M 2h[n]\cos(\omega n)$$

since  $h[n] = h[-n]$  for a zero phase filter.



This can be written as :

$$A(e^{jw}) = \sum_{k=0}^M a_k \cos(w)^k$$

Thus here we are expressing  $\cos(w)$  as a sum of powers of  $\cos(w)$  in the form:  $\cos(w) = T_n(\cos(w))$  where  $T_n(x)$  is an  $n^{\text{th}}$  order polynomial.  $a_k$ 's are constants that are related to  $h[n]$ .

We could re-write the equation

$$A(e^{jw}) = \sum_{k=0}^M a_k \cos(w)^k$$

as :

$$A(e^{jw}) = P(x) \Big|_{x=\cos w}$$

$$P(x) = \sum_{k=0}^M a_k x^k$$

- Thus  $A(e^{j\omega})$  can be expressed as a  $M^{\text{th}}$  degree polynomial.
- The key to arriving at an optimum design for the required filter is to fix  $\omega_s$ , and  $\omega_p$  and vary  $\delta_s$  and  $\delta_p$ .
- Parks and McClellan showed that with  $M$ ,  $\omega_s$ , and  $\omega_p$  fixed, the frequency selective filter design problem becomes a problem in Chebyshev approximation over disjoint sets.
- The error function is given by :  $E(\omega) = W(\omega)[H_d(e^{j\omega}) - A(e^{j\omega})]$  where  $W(\omega)$  is a weighting function which incorporates the approximation parameters into the design process.

## Optimum Design of Filters using Matlab

- The M-file function *remez* in the Signal Processing Toolbox of Matlab implements the Parks-McClellan Algorithm, which uses the Remez exchange algorithm and Chebyshev approximation to arrive at the desired filter.
- More about the Remez exchange algorithm could be found in “*Digital Filter Design*” by T.W. Parks and C.S Burrus.
- The filters designed are optimal in the sense that they minimize the maximum error between the desired and actual frequency responses. They are sometimes called *minmax* filters.

## Design example using Matlab

Specifications:

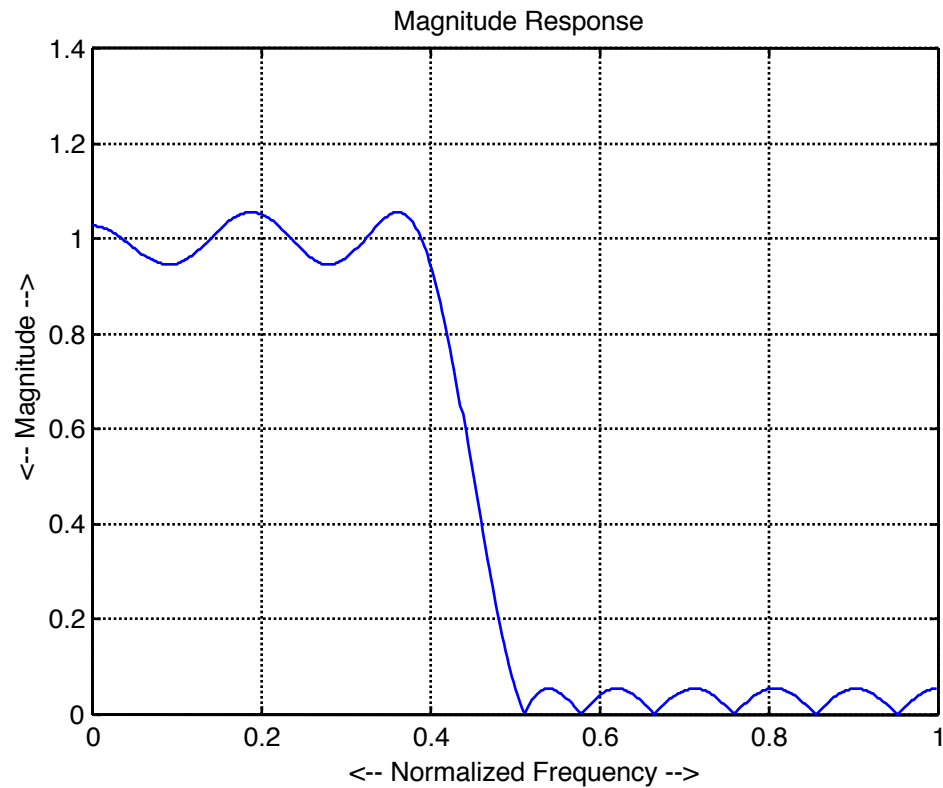
Filter Order = 20

Band Edges :  $w_p = 0.4$ ,  $w_s = 0.5$  {normalized frequencies}

Matlab Code:

```
n=20; %filter order
f=[0 0.4 0.5 1]; %frequency band edges
a=[1 1 0 0]; %desired amplitudes
coeff=remez(n, f, a);
[H,w]=freqz(coeff); %calculating the frequency response
plot(w/pi,abs(H)); %ploting the magnitude response
grid on;
xlabel('<-- Normalized Frequency -->');
ylabel('<-- Magnitude -->');
title('Magnitude Response');
```

## Magnitude Response of the Designed Filter :



## Other Matlab functions used for Optimal FIR filter Design

- Another very useful function from the Matlab Signal processing toolbox used for designing optimal FIR filters is the “*remezord*” function.
- This function gives the approximate order of the filter which we are designing based on the pass and stop band ripple and the cutoff frequencies.

Example:

Specifications of required filter:

Passband cutoff	= 500Hz
Stopband cutoff	= 600Hz
Sampling Frequency	= 2000Hz
Passband Ripple	= 3dB
Stopband Attenuation	= 40dB

## Matlab Code:

```
rp=3;           %passband ripple;
rs=40;         %stopband attenuation
Fs=2000;      %sampling frequency
f=[500 600];  %cutoff frequency
a=[1 0];      %desired amplitudes
d=[(10^(rp/20)-1)/(10^(rp/20)+1) 10^(-rs/20)];
[order, fo, ao, w]=remezord(f,a,d,Fs);
coeff=remez(order,fo,ao,w);

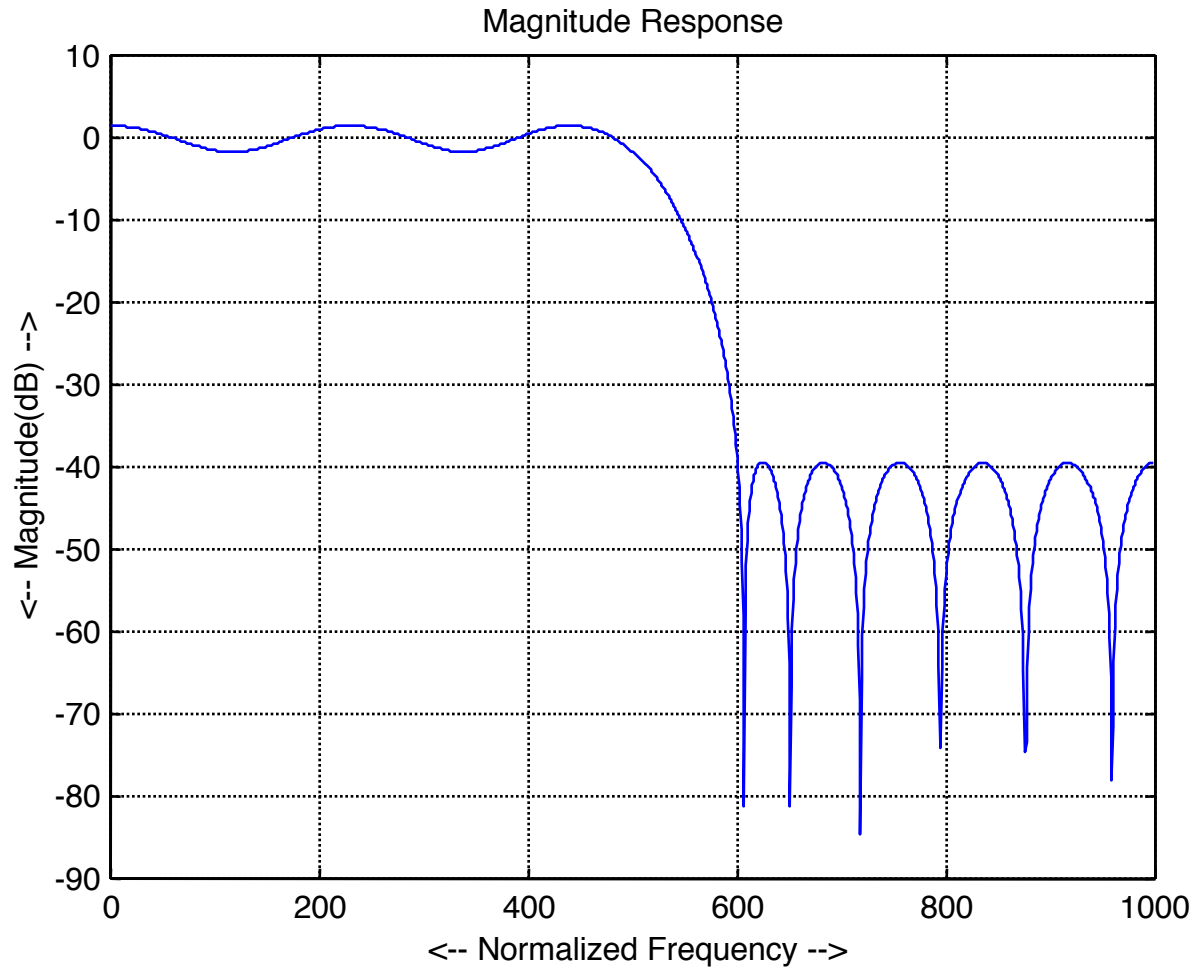
%calculating the frequency response
[H,w]=freqz(coeff,1,1024,Fs);

%plotting the magnitude response
plot(w,20*log10(abs(H)));
grid on;
xlabel('<-- Normalized Frequency -->');
ylabel('<-- Magnitude (dB) -->');
title('Magnitude Response');
```

# Design Example

contd..

## Magnitude Response of the Designed Filter :





# References

1. “Digital Signal Processing – A Practical Approach” -  
Emmanuel C. Ifeakor and Barrie W. Jervis  
Second Edition